

MT90866 **ZARLINK** Flexible 4 K x 2.4 K Channel Digital Switch with H.110 Interface and 2.4 K x 2.4 K Local Switch

Data Sheet

#### **Features**

- 2,432 x 2,432 non-blocking switching among local streams
- 4,096 x 2,432 blocking switching between backplane and local streams
- 2,048 x 2,048 non-blocking switching among ٠ backplane streams
- Rate conversion between backplane and local streams
- Rate conversion among local streams
- Backplane interface accepts data rates of 8.192 Mb/s or 16.384 Mb/s
- Local interface accepts data rates of 2.048 Mb/s, ٠ 4.096 Mb/s or 8.192 Mb/s
- Sub-rate switching (2 or 4 bits) configuration for local streams at a data rate of 2.048 Mb/s
- Meets all the key H.110 mandatory signal requirements including timing
- Per-channel variable or constant throughput ٠ delay
- Per-stream input delay, programmable for local streams on a per bit basis
- Per-stream output advancement, programmable for backplane and local streams
- Per-channel direction control for backplane streams
- Per-channel message mode for backplane and local streams
- Per-channel high impedance output control for backplane and local streams
- Compatible to Stratum 4 Enhanced clock switching standard
  - Integrated PLL conforms to Telcordia GR-1244-CORE Stratum 4 Enhanced switching standard -Holdover Mode with holdover frequency stability of 0.07 ppm
    - Jitter attenuation from 1.52 Hz.
    - Time interval error (TIE) correction
    - Master and Slave mode operation
- Non-multiplexed microprocessor interface

#### January 2006

#### **Ordering Information**

MT90866AG 344 Ball PBGA Trays MT90866AG2 344 Ball PBGA\* Trays \*Pb Free Tin/Silver/Copper

#### -40°C to +85°C

- Connection memory block-programming for fast device initialization
- Tristate-control outputs for external drivers
- Pseudo-Random Binary Sequence (PRBS) pattern generation and testing for backplane and local streams
- Conforms to the mandatory requirements of the IEEE-1149.1 (JTAG) standard
- 3.3 V operation with 5 V tolerant inputs and I/O's
- 5 V tolerant PCI driver on CT-Bus I/O's

# Applications

- Carrier-grade VoIP Gateways
- **IP-PBX and PABX**
- Integrated Access Devices
- Access Servers
- CTI Applications/CompactPCI® Platforms
- H.110, H.100, ST-BUS and proprietary Backplane Applications

#### Description

The MT90866 Digital Switch provides switching capacities of 4,096 x 2,432 channels between backplane and local streams, 2,432 x 2,432 channels among local streams and 2,048 x 2,048 channels among backplane streams. The local connected serial inputs and outputs have 32, 64 and 128 64 kb/s channels per frame with data rates of 2.048, 4.096 and 8.192 Mb/s respectively. The backplane connected serial inputs and outputs have 128 and 256 64 kb/s channels per frame with data rates of 8.192 and 16.384 Mb/s respectively.

Zarlink Semiconductor US Patent No. 5,602,884, UK Patent No. 0772912, France Brevete S.G.D.G. 0772912; Germany DBP No. 69502724.7-08

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The MT90866 also offers a sub-rate switching configuration which allows 2-bit wide 16 kb/s or 4-bit wide 32 kb/s data channels to be switched within the device.

The device has features that are programmable on a per-stream or a per-channel basis including message mode, input delay offset, output advancement offset, direction control, and high impedance output control.

The MT90866 supports all three of the H.110 specification required clocking modes: Primary Master, Secondary Master and Slave.



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# **Changes Summary**

The following table captures the changes from the September 2005 issue.

Page	Item	Change
9	Ball Signal Assignment	Added missing Ball signals.
11	Pin name: RESET	Worst case 600 $\mu$ s instead of 100 $\mu$ s is specified for the delay that must be applied before performing the first microprocessor access after the release of RESET pin.
44	Section 20.0	The 600 $\mu$ s delay requirement before performing the first microprocessor access after the release of RESET pin is added in this section for completeness.
80	Figure 40	Corrected 8 Mb/s stream channel to ch127

The following table captures the changes from the October 2003 issue.

Page	Item	Change
37, 58	Section 18.2 and Table 21	Added description clarifying that the MTIE reset must be set when the device is in the slave mode.
60	Table 22	Added MRST (bit 10) in MT90866 Mode Selection table
39, 41	Section 18.7 and Section 19.1	Deleted the intrinsic jitter descriptions in Section 18.7 and Section 19.1 and replaced them with "AC Electrical Characteristics <sup>†</sup> - Output Clock Jitter Generation (Unfiltered)" on page 76.
77, 78, 80	"AC Electrical Characteristics† - Backplane Serial Streams with Data Rate of 8 Mb/s", "AC Electrical Characteristics† - Backplane Serial Streams with Data Rate of 16 Mb/s" and "AC Electrical Characteristics† - Local Serial Stream Input Timing".	Input data sampling timings were updated for clarity purposes.

# MT90866

$\wedge$	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	
ب	STio	VDD5V	STio	STio 19	STio	STio 25	STio	STio 31	VDD5V	STi2	STi3	STi8	STi9	STi12	STi13	STi16	STi21	STi22	STi24	STi26	
A	14 STio	VDD5V	17 GND	19 STio					PCI_	STi2		STI6		GND							A
B	13		GND	18 STio 16	STio 21 STio 20	STio 24 STio 23	STio 27 STio 28	STio 30 STio 29	OE -	STI	STi5 STi4	STI6	STi10 STi11	GND	STi14	STi17	STi20	STi23	STi25	STi27	в
D	STio 12 STio	STio 15 STio 10	STio 11	16 GND	20 VDD	23 GND	28 VDD	29 GND	VDD	GND	VDD	VDD	GND	VDD	STi15 GND	STi18 VDD	STi19 GND	GND GND	GND STo0	GND STo1	C
E	STio 9 STio 6	10 STio 7	11 STio	VDD		_				-			-		-		VDD	LCSTo	ST02	STo3	E
F	6 STio 5	7 STio 2	8 STio 3	GND				GND	GND	GND	GND	GND	GND				GND	ST04	STo5	ST06	F
G	5 STio 4	2 STio 0	3 STio 1	VDD			GND	GND	GND	GND	GND	GND	GND	GND			VDD	STo8	STo9	STo7	G
н	4 VDD5V	0 GND	1 GND	GND		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		GND	STo10	STo11	STo13	н
J	DTA	BCSTo	D15	VDD		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		VDD		STo14	STo15	J
к	D12	D13	D14	VDD		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		GND	STo16	STo17	STo18	к
L	D9	D10	D11	GND		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		VDD	STo20	STo21	STo19	L
М	D6	D7	D8	VDD		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		VDD	STo22	STo23	STo24	м
Ν	D5	D3	D4	GND		GND	GND	GND	GND	GND	GND	GND	GND	GND	GND		GND	STo27	STo26	STo25	N
Р	D2	D1	D0	VDD			GND	GND	GND	GND	GND	GND	GND	GND		1	VDD	GND	B_Act ive	A_Act ive	Ρ
R	GND	GND	GND	GND		ļ		GND	GND	GND	GND	GND	GND		4		GND	ODE	C8_A_io	Frame_ A_io	R
т	A13	A12	A11	VDD													VDD	Frame_ B_io	C8_B_io	Fail_A	т
U	A10	A9	A8	GND	VDD	GND	VDD	GND	VDD	VDD	GND	VDD	GND	VDD	GND	VDD	GND	Fail_B	CTREF1	CTREF2	U
v	CS	A7	A5	A2	RESET	тск	IC2	IC4	IC6	IC3	TM1	GND	APLL VDD	APLL GND	LREF5	LREF2	NREFo	GND	ST CKō0	ST FP00	v
w	DS	A6	A3	A0	TMS	TDi	IC5	ST_ FP01	IC7	SG1	AT1	GND	C64 BYPS	FAIL_ PRI	LREF6	LREF3	LREF0	C32/64o	GND	GND	w
Y	R/W	A4	A1	TDo	TRST	IC0	IC1	ST_ CKo1	IC8	TM2	DT1	GND	C20i	FAIL_ SEC	LREF7	LREF4	LREF1	C1M5o	Sec_Los	Pri_Los	Y
	1	2	3	4	5	6	7	8	9 <b>—</b>	10	11	12	13	14	15	16	17	18	19	20	
Top View																					
1 - A1 corner is identified by metallized markings.																					

Figure 2 - 27 mm x 27 mm PBGA (JEDEC MO-151) Pinout

### **Ball Signal Assignment**

Ball	Signal										
A1	STIO14	C5	STIO20	F1	STIO5	H17	GND	L6	GND	N14	GND
A2	VDD5V	C6	STIO23	F2	STIO2	H18	STO10	L7	GND	N15	GND
A3	STIO17	C7	STIO28	F3	STIO3	H19	STO11	L8	GND	N17	GND
A4	STIO19	C8	STIO29	F4	GND	H20	STO13	L9	GND	N18	STO27
A5	STIO22	C9	GND	F8	GND	J1	DTA	L10	GND	N19	STO26
A6	STIO25	C10	STI0	F9	GND	J2	BCSTO	L11	GND	N20	STO25
A7	STIO26	C11	STI4	F10	GND	J3	D15	L12	GND	P1	D2
A8	STIO31	C12	STI7	F11	GND	J4	VDD	L13	GND	P2	D1
A9	VDD5V	C13	STI11	F12	GND	J6	GND	L14	GND	P3	D0
A10	STI2	C14	GND	F13	GND	J7	GND	L15	GND	P4	VDD
A11	STI3	C15	STI15	F17	GND	J8	GND	L17	VDD	P7	GND
A12	STI8	C16	STI18	F18	STO4	J9	GND	L18	STO20	P8	GND
A13	STI9	C17	STI19	F19	STO5	J10	GND	L19	STO21	P9	GND
A14	STI12	C18	GND	F20	STO6	J11	GND	L20	STO19	P10	GND
A15	STI13	C19	GND	G1	STIO4	J12	GND	M1	D6	P11	GND
A16	STI16	C20	GND	G2	STIO0	J13	GND	M2	D7	P12	GND
A17	STI21	D1	STIO9	G3	STIO1	J14	GND	M3	D8	P13	GND
A18	STI22	D2	STIO10	G4	VDD	J15	GND	M4	VDD	P14	GND
A19	STI24	D3	STIO11	G7	GND	J17	VDD	M6	GND	P17	VDD
A20	STI26	D4	GND	G8	GND	J18	STO12	M7	GND	P18	GND
B1	STIO13	D5	VDD	G9	GND	J19	STO14	M8	GND	P19	B_ACTIVE
B2	VDD5V	D6	GND	G10	GND	J20	STO15	M9	GND	P20	A_ACTIVE
B3	GND	D7	VDD	G11	GND	K1	D12	M10	GND	R1	GND
B4	STIO18	D8	GND	G12	GND	K2	D13	M11	GND	R2	GND
B5	STIO21	D9	VDD	G13	GND	K3	D14	M12	GND	R3	GND
B6	STIO24	D10	GND	G14	GND	K4	VDD	M13	GND	R4	GND
B7	STIO27	D11	VDD	G17	VDD	K6	GND	M14	GND	R8	GND
B8	STIO30	D12	VDD	G18	STO8	K7	GND	M15	GND	R9	GND
B9	PCI_OE	D13	GND	G19	STO9	K8	GND	M17	VDD	R10	GND
B10	STI1	D14	VDD	G20	STO7	K9	GND	M18	STO22	R11	GND
B11	STI5	D15	GND	H1	VDD5V	K10	GND	M19	STO23	R12	GND
B12	STI6	D16	VDD	H2	GND	K11	GND	M20	STO24	R13	GND
B13	STI10	D17	GND	H3	GND	K12	GND	N1	D5	R17	GND
B14	GND	D18	GND	H4	GND	K13	GND	N2	D3	R18	ODE
B15	STI14	D19	STO0	H6	GND	K14	GND	N3	D4	R19	C8_A_IO
B16	STI17	D20	STO1	H7	GND	K15	GND	N4	GND	R20	FRAME_A_IO
B17	STI20	E1	STIO6	H8	GND	K17	GND	N6	GND	T1	A13
B18	STI23	E2	STIO7	H9	GND	K18	STO16	N7	GND	T2	A12
B19	STI25	E3	STIO8	H10	GND	K19	STO17	N8	GND	Т3	A11
B20	STI27	E4	VDD	H11	GND	K20	STO18	N9	GND	T4	VDD
C1	STIO12	E17	VDD	H12	GND	L1	D9	N10	GND	T17	VDD
C2	STIO15	E18	LCSTO	H13	GND	L2	D10	N11	GND	T18	FRAME_B_IO
C3	GND	E19	STO2	H14	GND	L3	D11	N12	GND	T19	C8_B_IO
C4	STIO16	E20	STO3	H15	GND	L4	GND	N13	GND	T20	FAIL_A

Ball	Signal	Ball	Signal		
U1	A10	W6	TDi		
U2	A9	W7	IC5		
U3	A8	W8	ST_FPo1		
U4	GND	W9	IC7		
U5	VDD	W10	SG1		
U6	GND	W11	AT1		
U7	VDD	W12	GND		
U8	GND	W13	C64BYPS		
U9	VDD	W14	FAIL_PRI		
U10	VDD	W15	LREF6		
U11	GND	W16	LREF3		
U12	VDD	W17	LREF0		
U13	GND	W18	C32/640		
U14	VDD	W19	GND		
U15	GND	W20	GND		
U16	VDD	Y1	R/W		
U17	GND	Y2	A4		
U18	FAIL_B	Y3	A1		
U19	CTREF1	Y4	TDo		
U20	CTREF2	Y5	TRST		
V1	CS	Y6	IC0		
V2	A7	Y7	IC1		
V3	A5	Y8	ST_CKo1		
V4	A2	Y9	IC8		
V5	RESET	Y10	TM2		
V6	ТСК	Y11	DT1		
V7	IC2	Y12	GND		
V8	IC4	Y13	C20I		
V9	IC6	Y14	FAIL_SEC		
V10	IC3	Y15	LREF7		
V11	TM1	Y16	LREF4		
V12	GND	Y17	LREF1		
V13	APLLVDD	Y18	C1M5O		
V14	APLLGND	Y19	SEC_LOS		
V15	LREF5	Y20	PRI_LOS		
V16	LREF2				
V17	NREFo				
V18	GND				
V19	ST_CKo0				
V20	ST_FPo0				
W1	DS				
W2	A6				
W3	A3				
W4	A0				
W5	TMS	1	1		

### **Pin Description**

PBGA Ball Number	Name	Description
D5, D7, D9, D11, D12, D14, D16, E4, E17, G4, G17, J4, J17, K4, L17, M4, M17, P4, P17, T4, T17,U5, U7, U9, U10, U12, U14, U16	V <sub>DD</sub>	+3.3 Volt Power Supply.
A2,A9,B2,H1	V <sub>DD5V</sub>	<b>+5.0 V/+3.3 V Power Supply</b> . If 5 V power supply is tied to these pins, STio0-31 pins will meet 5 V PCI requirements. If 3.3 V power supply is tied to these pins, STio0-31 pins will meet 3.3 V PCI requirements.
B3, B14,C3, C9, C14, C18, C19, C20, D4, D6, D8, D10, D13, D15, D17, D18, F4, F8-F13, F17, G7-G14, H2, H3, H4, H6-H15, H17, J6-J15, K6-K15, K17, L4, L6-L15, M6-M15, N4, N6-15, N17, P7-P14, P18, R1, R2, R3, R4, R8-R13, R17, U4, U6, U8, U11, U13, U15, U17, V12, V18, W12, W19, W20, Y12	V <sub>SS</sub>	Ground.
V13	APLLV <sub>DD</sub>	<b>+3.3 Volt Analog PLL Power Supply</b> . No special filtering is required for this pin.
V14	APLLV <sub>ss</sub>	Analog PLL Ground
V5	RESET	<b>Device Reset (5 V Tolerant Input).</b> This input (active low) puts the device in its reset state; this state clears the device's internal counters and registers. To ensure proper reset action, the reset pin must be low for longer than 400 ns. To ensure proper operation, a delay of 600 $\mu$ s must be applied before the first microprocessor access is performed after the RESET pin is set high. The device reset also tristates ST00-27 and STi00-31, and sets the LCSTo and BCSTo pins. When in a RESET condition, the C8_A_io, FRAME_A_io, C8_B_io, and FRAME_B_io signals are tri-stated.
G2, G3, F2, F3, G1, F1, E1, E2, E3, D1, D2, D3, C1, B1, A1, C2	STio0-3, STio4-7, STio8-11, STio12-15	Serial Input/Output Streams 0 - 15 (5 V Tolerant PCI I/Os). In H.110 mode, these pins accept serial TDM data streams at 8.192 Mb/s with 128 channels per stream. In the 16 Mb/s mode, these pins accept serial TDM data streams at 16.384 Mb/s with 256 channels per stream respectively.

PBGA Ball Number	Name	Description
C4, A3, B4, A4, C5, B5, A5, C6, B6, A6, A7, B7, C7, C8, B8, A8	STio16 - 19, STio20 - 23, STio24 - 27, STio28 - 31	Serial Input/Output Streams 16 - 31 (5 V Tolerant PCI I/Os). In H.110 mode, these pins accept serial TDM data streams at 8.192 Mb/s with 128 channels per stream. In the 16 Mb/s mode, these pins are tristated internally and should be connected to ground.
C10, B10, A10, A11	STi0-3	Serial Input Streams 0 - 3 (5 V Tolerant Inputs). In 2 Mb/s, 4 Mb/s or 8Mb/s mode, these inputs accept data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these inputs accept a data rate of 2.048 Mb/s.
C11, B11, B12, C12	STi4 - 7	Serial Input Streams 4 - 7 (5 V Tolerant Inputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these inputs accept data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these inputs accept a data rate of 2.048 Mb/s.
A12, A13, B13, C13	STi8 - 11	Serial Input Streams 8 - 11 (5 V Tolerant Inputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these inputs accepts data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these inputs accept a data rate of 2.048 Mb/s.
A14, A15, B15, C15	STi12 - 15	Serial Input Streams 12 - 15 (5 V Tolerant Inputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these inputs accept data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these inputs accept a data rate of 2.048 Mb/s.
A16, B16, C16, C17, B17, A17, A18, B18, A19, B19, A20, B20	STi16 - 27	Serial Input Streams 16 - 27 (5 V Tolerant Inputs). In 2 Mb/s mode, these inputs accept data rates of 2.048 Mb/s with 32 channels per stream respectively. In 4 Mb/s or 8 Mb/s mode, the STi16 - 18 inputs accept data rates of 4.096 or 8.192 Mb/s with 64 or 128 channels per stream respectively. In 4 Mb/s or 8 Mb/s mode the STi19 - 27 inputs should be driven low. No sub-rate switching mode is offered for STi16-27.
D19, D20, E19, E20	STo0 - 3	Serial Output Streams 0 - 3 (5 V Tolerant Tri-State Outputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these outputs have data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these outputs have a data rate of 2.048 Mb/s.
F18, F19, F20, G20	STo4 - 7	Serial Output Streams 4 - 7 (5 V Tolerant Tri-State Outputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these outputs have data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these outputs have a data rate of 2.048 Mb/s.
G18, G19, H18, H19	STo8 - 11	Serial Output Streams 8 - 11 (5 V Tolerant Tri-State Outputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these outputs have data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these outputs have a data rate of 2.048 Mb/s.
J18, H20, J19, J20	STo12 - 15	Serial Output Streams 12 - 15 (5 V Tolerant Tri-State Outputs). In 2 Mb/s, 4 Mb/s or 8 Mb/s mode, these outputs have data rates of 2.048, 4.096 or 8.192 Mb/s with 32, 64 or 128 channels per stream respectively. In the 2-bit and 4-bit sub-rate modes, these outputs have a data rate of 2.048 Mb/s.

PBGA Ball Number	Name	Description
K18, K19, K20, L20, L18, L19, M18, M19, M20, N20, N19, N18	STo16 - 27	Serial Output Streams 16 to 27 (5 V Tolerant Tri-state Outputs). In 2 Mb/s mode, these outputs have data rate of 2.048 Mb/s with 32 channels per stream. In 4 Mb/s or 8 Mb/s mode, the STo16 - 18 outputs have data rates of 4.096 Mb/s or 8.192 Mb/s with 64 or 128 channels per stream respectively; STo19 - 27 are driven low. No sub-rate switching mode is offered for STo16-27.
R18	ODE	<b>Output Drive Enable (5 V Tolerant Input).</b> When this pin is low, STo0 to STo27, STio0 to STio31, C1M5o, C32/64o, ST_CKo0, ST_CKo1, ST_FPo0 and ST_FPo1 outputs are all in high-impedance state. When ODE is high all of the aforementioned pins are active.
J2	BCSTo	<b>Backplane Control Signal (Output)</b> . This pin is used for backplane external tristate controllers. When this signal is high, the corresponding output channels are in a high impedance state. BCSTo's bit rate is 32.768 MHz.
E18	LCSTo	<b>Local Control Signal (Output).</b> This pin is used for local external tristate control. When this signal is high, the corresponding ouput channels are in a high impedance state. The bit rate is 32.768 MHz.
Y13	C20i	Master Clock (5 V Tolerant Input). This pin accepts a 20.000 MHz clock.
R19	C8_A_io	Clock A (5 V Tolerant I/O). This is a 8.192 MHz clock with 50% duty cycle.
R20	FRAME_A_io	Frame Reference A (5 V Tolerant I/O). This is a 122 ns wide, negative pulse, with 125 us period.
P20	A_Active	A Clock Active Indicator (5 V Tolerant Output): This pin indicates whether the C8_A_io and the FRAME_A_io pins are inputs or outputs. When Bit 13 of the DOM1 register is low, this pin drives low and the C8_A_io and FRAME_A_io output drivers are disabled. When Bit 13 of the DOM1 register is high, this pin drives high and the C8_A_io and FRAME_A_io output drivers are enabled.
T19	C8_B_io	Clock B (5 V Tolerant I/O). This is a 8.192 MHz clock with 50% duty cycle.
T18	FRAME_B_io	Frame Reference B (5 V Tolerant I/O). This is a 122 ns wide, negative pulse, with 125 us period.
P19	B_Active	<b>B Clock Active Indicator (5 V Tolerant Output):</b> This pin indicates whether the C8_B_io and the FRAME_B_io pins are inputs or outputs. When Bit 14 of the DOM1 register is low, this pin drives low and the C8_B_io and FRAME_B_io output drivers are disabled. When Bit 14 of the DOM1 register is high, this pins drives high and the C8_B_io and FRAME_B_io output drivers are enabled.
T20	FAIL_A	<b>A Failure (Output).</b> When the C8_A_io or the FRAME_A_io signal fails, this signal goes to high.
U18	FAIL_B	<b>B Failure (Output).</b> When the C8_B_io or the FRAME_B_io signal fails, this signal goes to high.

PBGA Ball Number	Name	Description			
U19	CTREF1	<b>CT-Bus Reference 1 (5 V Tolerant Input).</b> This pin accepts 8 KHz, 1.544 MHz or 2.048 MHz network timing reference.			
U20	CTREF2	<b>CT-Bus Reference 2 (5 V Tolerant Input).</b> This pin accepts 8 KHz, 1.544 MHz or 2.048 MHz network timing reference.			
W17, Y17, V16, W16, Y16, V15, W15, Y15	LREF0- 7	<b>Local Reference (5 V Tolerant Inputs).</b> These pins accept 8 KHz, 1.544 MHz or 2.048 MHz local timing reference.			
V17	NREFo	<b>Network Reference Output (Output).</b> Any local reference can be switched to this output. The output data rate can be either the same as the selected reference input data rate or divided to be 8 KHz.			
Y20	PRI_LOS	<b>Primary Reference Lost (5 V Tolerant Input).</b> When this signal is high, it indicates that PRIMARY REFERENCE is not valid. Combined with SEC_LOS input, this input pin is used in the External Reference Switching Mode of the DPLL.			
Y19	SEC_LOS	Secondary Reference Lost (5 V Tolerant Input). When this signal is high, it indicates that SECONDARY REFERENCE is not valid. Combined with the PRI_LOS input, this input pin is used in the External Reference Switching Mode of the DPLL.			
W14	FAIL_PRI	<b>Primary Reference Failure (5 V Tolerant Output)</b> . This pin reflects the logic status of the PLS bit of the DPLL House Keeping Register (DHKR). When the primary reference fails, this signal goes to 1.			
Y14	FAIL_SEC	Secondary Reference Failure (5 V Tolerant Output). This pin reflects the logic status of the SLS bit of the DPLL House Keeping Register (DHKR). When the secondary reference fails, this signal goes to 1.			
W18	C32/64o	<b>32/64o Clock (5 V Tolerant Output).</b> A 32.768 MHz output clock when the PLL Clock Monitor register bit (CKM) is low. A 65.536 MHz clock when the PLL Clock Monitor register bit (CKM) is high.			
Y18	C1M5o	C1.50 Clock (5 V Tolerant Output). A 1.544 MHz output clock.			
V20	ST_FPo0	<b>ST-Bus Frame Pulse Output (5 V Tolerant Output).</b> The width of this output ST-Bus frame pulse can be 244 ns, 122 ns or 61 ns. The frequency is 8 KHz.			
V19	ST_CKo0	<b>ST-Bus Clock Output (5 V Tolerant Output).</b> The frequency of this output ST-Bus clock can be 4.096 MHz, 8.192 MHz or 16.384 MHz.			
W8	ST_FPo1	<b>ST-Bus Frame Pulse Output (5 V Tolerant Output)</b> . The width of this output ST-Bus frame pulse can be 244 ns, 122 ns or 61 ns. The frequency is 8 KHz.			
¥8	ST_CKo1	<b>ST-Bus Clock Output (5 V Tolerant Output)</b> . The frequency of this output ST-Bus clock can be 4.096 MHz, 8.192 MHz or 16.384 MHz.			
V1	CS	Chip Select (5 V Tolerant Input). This active low input is used by the microprocessor to access the microport.			

PBGA Ball Number	Name	Description			
W1	DS	Data Strobe (5 V Tolerant Input). This active low input works in conjunction with CS to initiate the read and write cycles.			
Y1	R/W	<b>Read/Write (5 V Tolerant Input).</b> This input controls the direction of the data bus lines (D0 - D15) during the microprocessor access.			
W4, Y3, V4, W3, Y2, V3, W2, V2, U3, U2, U1, T3, T2, T1	A0 - A13	Address 0 - 13 (5 V Tolerant Inputs). These are the address lines to the internal memories and registers.			
P3, P2, P1, N2, N3, N1, M1, M2, M3, L1, L2, L3, K1, K2, K3, J3	D0 - D15	Data Bus 0 - 15 (5 V Tolerant I/Os). These pins form the 16-bit data bus of the microport.			
J1	DTA	<b>Data Transfer Acknowledge (5 V Tolerant Output)</b> . This active low output indicates that a data bus transfer is completed. A pull-up resistor is required to hold a high level.			
B9	PCI_OE	<b>PCI Output Enable (3.3 V Tolerant Input)</b> . This active low input is the control signal used to tristate the STio0 - 31 pins during hot-swapping. During normal operation this signal should be low.			
W13	C64BYPS	PLL Bypass Clock Input (5 V Tolerant Input). Used for device testing. In functional mode, this input MUST be low.			
V11	TM1	<b>APLL Test Pin 1 (3.3 V Input).</b> Use for APLL testing only. In normal operation, this input should be connected to ground.			
Y10	TM2	<b>APLL Test Pin 2 (3.3 V Input).</b> Use for APLL testing only. In normal operation, this input should be connected to ground.			
W10	SG1	PLL Test Control (3.3 V Input). Use for APLL testing only. In normal peration, this input should be connected to ground.			
W11	AT1	Analog Test Access (5 V Tolerant I/O). Use for APLL testing only. No connection for normal operation.			
Y11	DT1	<b>Digital Test Access Output (5 V Tolerant Output).</b> Use for APLL testing only. No connection for normal operation.			
W5	TMS	<b>Test Mode Select (3.3 V Input with Internal pull-up).</b> JTAG signal that controls the state transitions of the TAP controller. This pin is pulled high by an internal pull-up when not driven.			
W6	TDi	<b>Test Serial Data In (3.3 V Input with Internal pull-up)</b> . JTAG serial test instructions and data are shifted in on this pin. This pin is pulled high by an internal pull-up when not driven.			
Y4	TDo	<b>Test Serial Data Out (3.3 V Tolerant Tri-state Output)</b> . JTAG serial data is output on this pin on the falling edge of TCK. This pin is held in high impedance state when JTAG is not enabled.			

PBGA Ball Number	Name	Description
V6	ТСК	<b>Test Clock (5 V Tolerant Input).</b> Provides the clock to the JTAG test logic. This pin should be low when JTAG is not enabled.
Y5	TRST	<b>Test Reset (3.3 V Input with Internal pull-up).</b> Asynchronously initializes the JTAG TAP Controller by putting it in the Test-Logic-Reset state. This pin should be pulled low to ensure that the MT90866 is in normal functional mode.
Y6	IC0	Leave unconnected for normal operation.
Y7	IC1	Leave unconnected for normal operation.
V7	IC2	In normal operation this pin MUST be connected to ground.
V10	IC3	Leave unconnected for normal operation.
V8	IC4	Leave unconnected for normal operation.
W7	IC5	Leave unconnected for normal operation.
V9	IC6	Leave unconnected for normal operation.
W9	IC7	Leave unconnected for normal operation.
Y9	IC8	Leave unconnected for normal operation.

# 1.0 Device Overview

The MT90866 can switch up to  $4,096 \times 2,432$  channels while providing a rate conversion capability. It is designed to switch 64 kb/s PCM or N X 64 kb/s data between the backplane and local switching applications. The device maintains frame integrity in data applications and minimum throughput delay for voice application on a per channel basis.

The backplane interface can operate at 8.192 Mb/s in CT-Bus mode or 16.384 Mb/s in ST-BUS mode and is arranged in 125  $\mu$ s wide frames that contain 128 or 256 channels respectively. A built-in rate conversion circuit allows users to interface between backplane and local interfaces which operates at 2.048 Mb/s, 4.096 Mb/s or 8.192 Mb/s. When the device is in the local sub-rate switching mode, 2-bit 16 kb/s or 4-bit 32 kb/s data channels can be switched within the device. The local sub-rate switching mode is available in 2 Mb/s mode only.

By using Zarlink's message mode capability, the microprocessor can access input and output time slots on a per channel basis. This feature is useful for transferring control and status information for external circuits or other TDM devices.

# 2.0 Functional Description

A Functional Block Diagram of the MT90866 is shown in Figure 1, "Functional Block Diagram" on page 2. It is designed to interface CT-Bus and ST-BUS serial streams from a backplane source and ST-BUS serial streams from a local source.

# 3.0 Frame Alignment Timing

In the ST-BUS or the CT-Bus mode, the C8\_A\_io or C8\_B\_io pin accepts an 8.192 MHz clock for the frame pulse alignment. The FRAME\_A\_io or FRAME\_B\_io is the frame pulse signal which goes low at the frame boundary for 122 ns. The frame boundary is defined by the rising edge of the C8\_A\_io or C8\_B\_io clock during the low cycle of the frame pulse. Figure 3, "CT-Bus Timing for 8 Mb/s Backplane Data Streams" on page 17 is the CT-Bus timing for the backplane 8.192 Mb/s data streams and Figure 4, "ST-BUS Timing for 16 Mb/s Backplane Data Streams" on page 18 is the ST-BUS timing for the 16.384 Mb/s backplane data stream.



Figure 3 - CT-Bus Timing for 8 Mb/s Backplane Data Streams



Figure 4 - ST-BUS Timing for 16 Mb/s Backplane Data Streams

# 4.0 Switching Configuration

The device has two operation modes at different data rates for the backplane interface and five operation modes for the local interface. These modes can be programmed via the Device Mode Selection (DMS) register. Mode selections between the backplane and local interfaces are independent.

#### 4.1 Backplane Interface

The backplane interface can be programmed to accept data streams of 8 Mb/s or 16 Mb/s. When H.110 mode is enabled, STio0 to STio31 have a data rate of 8.192 Mb/s. When ST-BUS mode is enabled, STio0 to STio15 have a data rate of 16.384 Mb/s. Table 1 on page 18 describes the data rates and mode selections for the backplane interface.

#### 4.2 Local Interface

Five operation modes, 2 Mb/s, 4 Mb/s, 8 Mb/s, 2-bit sub-rate and 4-bit sub-rate switching, can be selected for the local ST-BUS interface. The local interface is divided into five groups. Group 0 contains STi/STo0-3, Group 1 contains STi/STo4-7, Group 2 contains STi/STo8-11, Group 3 contains STi/STo12-15 and Group 4 contains STi/STo16-27. Each group can be selected individually through the Device Mode Selection (DMS) register. Streams belonging to the same group have the same operation mode. For Groups 0 to 3, any one of the five operation modes can be selected. Input data streams STi0-15 and output data streams, STo0 -15 can be selected according to the group to which they belong. STi16-27 and output data streams STo16-27 belong to Group 4 and can operate in 2 Mb/s mode. In Group 4, only input data streams Sti16-18 and output data streams Sto16-18 can operate in 4 Mb/s and 8 Mb/s mode. When operating Group 4 at 4 Mb/s or 8 Mb/s the unused output streams, STo19-27 are driven low. No sub-rate modes are available for Group 4 data streams. See Table 2 on page 19 to Table 6 on page 20 for a description of the data rates and mode selection for the local ST-BUS interface.

BMS bit of the DMS Register	Modes	Backplane Interface
0	8.192 Mb/s	STio0 - 31
1	16.384 Mb/s	STio0 - 15

#### Table 1 - Mode Selection for Backplane Streams

ſ	OMS Register Bit	S	Modes	Usable Streams
LG02	LG02 LG01		Modes	Usable Streams
0	0	0	8.192 Mb/s	
0	0	1	4.096 Mb/s	STi0 - 3, STo0 - 3
0	1	0	2.048 Mb/s	
0	1	1	4-bit subrate	
1	0	0	2-bit subrate	

Table 2 - Mode Selection for Local STi0 - 3 and STo0 - 3 Streams, Group 0

ſ	DMS Register Bits	s	Madaa	llaabla Streama
LG12	LG11	LG10	Modes	Usable Streams
0	0	0	8.192 Mb/s	
0	0	1	4.096 Mb/s	
0	1	0	2.048 Mb/s	
0	1	1	4-bit subrate	STi4 - 7, STo4 -7
1	0	0	2-bit subrate	

Table 3 - Mode Selection for Local STi4 - 7 and STo4 - 7 Streams, Group 1

I	DMS Register Bits	3	Madaa	Usable Streams
LG22	LG21 LG20		Modes	Usable Streams
0	0	0	8.192 Mb/s	
0	0	1	4.096 Mb/s	STi8 - 11, STo8 - 11
0	1	0	2.048 Mb/s	
0	1	1	4-bit subrate	
1	0	0	2-bit subrate	

Table 4 - Mode Selection for Local STi8 - 11 and STo8 - 11 Streams, Group 2

	DMS Register Bits	;	Madaa	Hackle Streams
LG32	LG31	LG30	Modes	Usable Streams
0	0	0	8.192 Mb/s	
0	0	1	4.096 Mb/s	STi12-15, STo12-15
0	1	0	2.048 Mb/s	
0	1	1	4-bit subrate	
1	0	0	2-bit subrate	

Table 5 - Mode Selection for Local STi12 - 15 and ST012 - 15 Streams, Group 3

DMS Reg	jister Bits	Madaa	Usable Streams		
LG41	LG40	Modes			
0	0	8.192 Mb/s	STi16 - 18, STo16 - 18		
0	1	4.096 Mb/s			
1	0	2.048 Mb/s	STi16 - 27, STo16 - 27		

Table 6 - Mode Selection for Local STi16 - 27 and STo16 - 27 Streams, Group 4

# 5.0 Local Input Delay Selection

The local input delay selection allows individual local input streams to be aligned and shifted against the input frame pulse (FRAME\_A\_io or FRAME\_B\_io). This feature compensates for the variable path delays in the local interface. Such delays can occur in large centralized and distributed switching system.

Each local input stream can have its own bit delay offset value by programming the local input bit delay selection registers (LIDR0 to LIDR9). See Table 12, "Local Input Bit Delay Registers (LIDR0 to LIDR9) Bits" on page 50, for the contents of these registers. Possible bit adjustment can range up to +7 3/4 bit periods forward with resolution of 1/4 bit period. See Table 13 on page 51 and Figure 19 on page 51 for local input delay programming.

# 6.0 Output Advancement Selection

The MT90866 allows users to advance individual backplane or local output streams with respect to the frame boundary. This feature is useful in compensating variable output delays caused by various output loading conditions. Each output stream can have its own advancement value programmed by the output advancement registers. The backplane output advancement registers (BOAR0 to BOAR3) are used to program the backplane output advancement. The local output advancement registers (LOAR0 to LOAR3) are used to program the local output advancement. Possible adjustment for local and backplane output data streams is 22.5 ns with a resolution of 7.5 ns. The advancement is independent of the output data rate. Table 14 on page 52 and Figure 20, "Example of Backplane Output Advancement Timing" on page 52, and Table 15 on page 53 and Figure 21, "Local Output Advancement Timing" on page 52 and Figure 21, "Local Output Advancement Timing" on page 53 describe the details of the output advancement programming for the backplane and local interfaces respectively.

# 7.0 Local Output Timing Considerations

The output data of the MT90866's local side is slightly advanced with respect to the frame and bit boundary as defined by the local output clocks and frame pulses (ST\_FPo0, ST\_CKo0, ST\_FPo1, ST\_CKo1). The advancement is in the range of 5 ns to 17 ns. Despite this advancement, the MT90866 will operate within the parameters specified in the datasheet because input data are usually sampled at the 3/4 or 1/2 point of the bit cell. However, the user should be cautious when introducing additional delay to the clock signals only (e.g., by passing them through glue logic, FPGA, or CPLD), which will introduce a few nanoseconds of delay relative to the data. If the clock signal is delayed, data will be advanced from the receiver device's point of view. This may cause errors in sampling the data. Using an example where a 3/4 sampling point is used, there is about 30 ns from the sampling point to the end of the bit cell. With a worst-case of 17 ns advancement, the timing margin will be approximately 13 ns. Any additional delays applied to the local output clocks (ST\_CKo0 and ST\_CKo1) must not exceed 13 ns minus the hold time of the receiving device. Delays applied to both clocks and data equally will not impact the device operation.

# 8.0 Memory Block Programming

The MT90866 block programming mode (BPM) register provides users with the capability of initializing the local and backplane connection memories in two frames. The local connection memory is partitioned into high and low parts. Bit 13 - bit 15 of every backplane connection memory location will be programmed with the pattern stored in bit 6 - bit 8 of the BPM register. Bit 13 - bit 15 of every local connection memory low location will be programmed with the pattern stored in bits 3 to 5 of the BPM register. The other bit positions of the backplane connection memory, the local low connection memory and all bits of the local high connection memory are loaded with zeros. See Figure 5, "Block Programming Data in the Connection Memories" on page 22 for the connection memory contents when the device is in block programming mode.

The block programming mode is enabled by setting the memory block program (MBP) bit of the Control register to high. After the block programming enable (BPE) bit of the BPM register is set to high, the block programming data will be loaded into bits 13 to 15 of every backplane connection memory location and bits 13 to 15 of every local connection memory low location. The other connection memory bits are loaded with zeros. When the memory block programming is completed, the device resets the BPE bit to low. See Table 11 on page 49 for the bit assignment of the BPM register.

# 9.0 Delay Through the MT90866

The switching of information from the input serial streams to the output serial streams results in a throughput delay. The device can be programmed to perform time slot interchange functions with different throughput delay capabilities on a per-channel basis. For voice applications it is recommended to select variable throughput delay to ensure minimum delay between input and output data. In wideband data applications it is recommended to select constant throughput delay to maintain the frame integrity of the information through the switch.

The delay through the device varies according to the type of throughput delay selected in the BTM2 - BTM0 bits of the backplane connection memory or LTM0 - LTM2 bits of the local connection memory as described in Table 25 on page 63 and Table 29 on page 65, respectively.

#### 9.1 Variable Delay Mode

The delay in this mode is dependent only on the combination of source and destination channels and is independent of input and output streams. The minimum delays achievable in the MT90866 device are 3-channel delay, 5-channel delay, and 10-channel delay for the 2 MB/s, 4 MB/s, and 8 MB/s respectively. The maximum delay is one frame plus three channels, one frame plus five channels, and one frame plus ten channels for the 2 Mb/s, 4 Mb/s and 8 Mb/s modes respectively.

For the backplane interface, the variable delay mode can be programmed through the backplane connection memory bits, BTM2 - BTM0. When BTM2 - BTM0 are programmed to "000", it is a per-channel variable delay from

local input to the backplane output. When BTM2 - BTM0 are set to "010", it is a per-channel variable delay from backplane input to backplane output.

For the local interface, the variable delay mode can be programmed through the local connection memory low bits, LTM2 - LTM0. When LTM2 - LTM0 is programmed to "000", it is a per-channel variable delay from local input to local output. When LTM2 - LTM0 is set to "010", it is a per-channel variable delay from backplane input to local output.



Figure 5 - Block Programming Data in the Connection Memories

#### 9.2 Constant Delay Mode

In this mode, a multiple data memory buffer is used to maintain frame integrity in all switching configurations by using three pages of Data Memory where a channel written in any of the buffers during frame N is always read out during frame N+2.

For the backplane interface, when BTM2 - BTM0 is programmed to "001", it is a per-channel constant delay mode from local input to backplane output. When BTM2 - BTM0 is programmed to "011", it is a per-channel constant delay from backplane input to backplane output.

For the local interface, when LTM2 - LTM0 is programmed to "001", it is a per-channel constant delay mode from local input to local output. When LTM2 - LTM0 is set to "011", it is a per-channel constant delay mode from backplane input to local output.

# **10.0** Microprocessor Interface

The MT90866 provides a parallel microprocessor interface for non-multiplexed bus structures. This interface is compatible with Motorola non-multiplexed bus structure. The required microprocessor signals are the 16-bit data bus (D15-D0), 14-bit address bus (A13-A0) and 4 control lines ( $\overline{CS}$ ,  $\overline{DS}$ , R/W and  $\overline{DTA}$ ). See Figure 44, "Motorola Non-Multiplexed Bus Timing" on page 83 for the Motorola non-multiplexed bus timing.

The MT90866 microprocessor port provides access to the internal registers, the connection and data memories. All locations provide read/write access except for the Local and Backplane Bit Error Rate registers (LBERR and BBERR) and Data Memory which can only be read by the users.

# 10.1 DTA Data Transfer Acknowledgment Pin

The  $\overline{\text{DTA}}$  pin of the microprocessor is driven LOW by internal logic to indicate that a data bus transfer is completed. When the bus cycle ends, this pin switches to the high impedance state. An external pull-up of between 1 K $\Omega$  and 10 K $\Omega$  is required at this output.

# 11.0 Address Mapping of Memories and Registers

The address bus on the microprocessor interface selects the internal registers and memories of the MT90866. If the address bit A13 is low, then the registers are addressed by A12 to A0 as shown in Table 7 on page 23.

A13 - A0	Location
0000 <sub>H</sub>	Control Register, CR
0001 <sub>H</sub>	Device Mode Selection Register, DMS
0002 <sub>H</sub>	Block Programming Mode Register, BPM
0003 <sub>H</sub>	Reserved
0004 <sub>H</sub>	Local Input Bit Delay Register 0, LIDR0
0005 <sub>H</sub>	Local Input Bit Delay Register 0, LIDR1
0006 <sub>H</sub>	Local Input Bit Delay Register 2, LIDR2
0007 <sub>H</sub>	Local Input Bit Delay Register 3, LIDR3
0008 <sub>H</sub>	Local Input Bit Delay Register 4, LIDR4
0009 <sub>H</sub>	Local Input Bit Delay Register 5, LIDR5
000A <sub>H</sub>	Local Input Bit Delay Register 6, LIDR6
000B <sub>H</sub>	Local Input Bit Delay Register 7, LIDR7
000C <sub>H</sub>	Local Input Bit Delay Register 8, LIDR8
000D <sub>H</sub>	Local Input Bit Delay Register 9, LIDR9
000E <sub>H</sub> to 001B <sub>H</sub>	Reserved
001C <sub>H</sub>	Backplane Output Advancement Register 0, BOAR0
001D <sub>H</sub>	Backplane Output Advancement Register 1, BOAR1
001E <sub>H</sub>	Backplane Output Advancement Register 2, BOAR2
001F <sub>H</sub>	Backplane Output Advancement Register 3, BOAR3
0020 <sub>H</sub>	Local Output Advancement Register 0, LOAR0
0021 <sub>H</sub>	Local Output Advancement Register 1, LOAR1
0022 <sub>H</sub>	Local Output Advancement Register 2, LOAR2
0023 <sub>H</sub>	Local Output Advancement Register 3, LOAR3
0024 <sub>H</sub> to 0026 <sub>H</sub>	Reserved
0027 <sub>H</sub>	Local BER Input Selection Register, LBIS
0028 <sub>H</sub>	Local BER Register, LBERR
0029 <sub>H</sub>	Backplane BER Input Selection Register, BBIS
002A <sub>H</sub>	Backplane BER Register, BBERR
002B <sub>H</sub>	DPLL Operation Mode Register 1, DOM1

Table 7 - Address Map For Internal Registers (A13 = 0)

A13 - A0	Location
002C <sub>H</sub>	DPLL Operation Mode Register 2, DOM2
002D <sub>H</sub>	DPLL Output Adjustment Register, DPOA
002E <sub>H</sub>	DPLL House Keeping Register, DHKR

#### Table 7 - Address Map For Internal Registers (A13 = 0) (continued)

If A13 is high, the remaining address input lines are used to select the data and connection memory positions corresponding to the serial input or output data streams as shown in Table 8 on page 25.

The Control register (CR), the Device Mode Selection register (DMS) and the Block Programming Mode register (BPM) control all the major functions of the device. The DMS and BPM should be programmed immediately after system power up to establish the desired switching configuration as explained in the Frame Alignment Timing and Switching Configurations sections. The Control register is used to select Data or Connection Memory for microport operations, ST-BUS output frame and clock modes, and to set Memory Block Programing and Bit Error Rate Testing.

The Control register (CR) consists of the memory block programming bit (MBP) and the memory select bits (MS2-0). The memory block programming bit allows users to program the entire connection memory in two frames (see Memory Block Programming section). The memory select bits control the selection of the connection memories or the data memories. See Table 9 on page 46 for content of the Control register.

The DMS register consists of the backplane and the local mode selection bits (BMS, LG41 - LG40, LG32 - LG30, LG22 - LG20, LG12 - LG10 and LG02 - LG00) that are used to enable various switching modes for the backplane and the local interfaces respectively. See Table 10 on page 47 for the content of the DMS register.

The BPM register consists of the block programming data bits (LBPD2-0 and BBPD2-0) and the block programming enable bit (BPE). The block programming enable bit allows users to program the entire backplane and local connection memories in two frames (see Memory Block Programming section). If the ODE pin is low, the backplane CT-Bus is in input mode and the local output drivers are in high impedance state. If the ODE pin is high, all the backplane CT-Bus and local ST-BUS output drivers are controlled on a per channel basis by backplane and local connection memories, respectively. By programming BTM2 through BTM0 bits to "110" in the backplane connection memory, the user can control the per-channel input on the backplane interface. For the local interface, users can program LTM2 -0 bits to "110" in the local connection memory to control the per-channel high impedance output on the local ST-BUS. See Table 11 on page 49 for the content of the BPM register.

2 A11 0 0 0 0 0 0 0 0	A10 0 0 0 0	A9 0 0 1	A8 0 1 0	Stream # Stream 0 Stream 1	A7 0 0	A6 0	A5 0	A4 0	A3 0	A2 0	A1 0	A0 0	Channel # Ch 0
0 0 0 0	0 0 0	0	1		-	-	-	-	0	0	0	0	Ch 0
0 0 0	0	-	•	Stream 1	0	~	-						
0	0	1	Δ			0	0	0	0	0	0	1	Ch 1
0	-		0	Stream 2	-	-							
•	- 1	1	1	Stream 3	-								
0		0	0	Stream 4	0	0	0	1	1	1	1	0	Ch 30
	1	0	1	Stream 5	0	0	0	1	1	1	1	1	Ch 31 (Note 2)
0	1	1	0	Stream 6	0	0	1	0	0	0	0	0	Ch 32
0	1	1	1	Stream 7	0	0	1	0	0	0	0	1	Ch 33
1	0	0	0	Stream 8	-	-							
					-								
					0	0	1	1	1	1	1	0	Ch 62
					0	0	1	1	1	1	1	1	Ch 63 (Note 3 & 6)
	-	-	•		-	-							
		-			-	-					•		
		-			0	1	1	1	1	1	1	0	Ch 126
1	0	1	1	Stream 27	0	1	1	1	1	1	1	1	Ch 127 (Note 4 & 7)
1	1	0	0	Stream 28	-	-							
1	1	0	1	Stream 29		-		-					
1	1	1	0	Stream 30	1	1	1	1	1	1	1	0	Ch 254
1	1	1	1	Stream 31	1	1	1	1	1	1	1	1	Ch 255 (Note 5)
С			.         .         .           .         .         .           .         .         .           .         .         .           .         .         .           .         .         .           .         .         .           1         0         1           1         1         0           1         1         1           1         1         1           1         1         1           1         1         1           1         1         1	.     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       .     .     .     .       1     0     1     1       1     1     0     1       1     1     1     1       0     1     1     1       1     1     1     1	.     .     .     .     .       .     .     .     .     .       .     .     .     .     .       .     .     .     .     .       .     .     .     .     .       .     .     .     .     .       .     .     .     .     .       .     .     .     .     .       1     0     1     1     Stream 28       1     1     0     1     Stream 30       1     1     1     1     Stream 31	.       .	.       . <td.< td=""> <td.< td=""> <td.< td=""></td.<></td.<></td.<>	.       .	.       .	.       .			1       0       0       0       0       1



# 12.0 Backplane Connection Memory

The backplane connection memory controls the switching configuration of the backplane interface. Locations in the backplane connection memory are associated with particular STio streams.

The BTM2 - 0 bits of each backplane connection memory allows the per-channel selection for the message or the connection mode, the constant or the variable delay mode, the high impedance control of the STio driver or the bit error test enable. See Table 25 on page 63 for the content per-channel control function.

In the switching mode, the contents of the backplane connection memory stream address bits (BSAB4-0) and channel address bits (BCAB7-0) define the source information (stream and channel) of the time slot that will be switched to the backplane STio streams. During the message mode, only the lower 8 bits (8 least significant bits) of the backplane connection memory will be transferred to the STio pins.

# 13.0 Local Connection Memory

The local connection memory controls the local interface switching configuration. Local connection memory is split into a high and a low part. Locations in the local connection memory are associated with particular STo output streams.

The LTM2 - 0 bits of each local connection memory low allows the per-channel selection for the message or the connection mode, the constant or the variable delay mode, the high impedance control of the STo driver or the bit error test enable. See Table 29 on page 65 for the content per-channel control function.

In the switching mode, the contents of the local connection memory low stream address bits (LSAB4-0) and the channel address bits (LCAB7-0) of the local connection memory defines the source information (stream and

channel) of the time slot that will be switched to the local STo streams. During the message mode, only the lower 8 bits (8 least significant bits) of the local connection memory low bits are transferred to the STo pins.

In the sub-rate switching mode, although the output channels are divided up into 2 or 4-bit channels, the input streams still have 8-bit channel boundaries. Therefore, it is necessary to indicate which bits in the input 8-bit channel will be switched out to the 2 or 4-bit channel. When 2-bit or 4-bit sub-rate switching is enabled, the LSRS1-0 bits in the local connection memory high define which bit positions contains the sub-rate data.

#### 14.0 Bit Error Rate Test

The MT90866 offers users a Bit Error Rate (BER) test feature for the backplane and the local interfaces. The circuitry of the BER test consists of a transmitter and a receiver on both interfaces that can transmit and receive the BER patterns independently. The transmitter can output a pseudo random patterns of the form 2<sup>15</sup> - 1 to any channel and any stream within a frame time. For the test, users can program the output channel and stream through the backplane or local connection memory and the input channel and stream using Local or Backplane BER Input Selection (BIS) registers. See Table 16 on page 54 and Table 18 on page 54 for the LBIS and the BBIS registers contents, respectively.

The receiver receives the BER pattern and does an internal BER pattern comparison. For backplane interface, the comparison result is stored in the Backplane BER register (BBERR). For local interface, the result is stored in the Local BER register (LBERR).

#### 15.0 External Tristate Control

The MT90866 has the flexibility to provide users with the choice of external per-channel tristate control. Two control signals are provided. For the backplane interface, it is the BCSTo output. For the local interface, it is the LCSTo output. Each control signal has a data rate of 32.768 Mb/s with 4,096 control bits per frame. Each bit position corresponds to a specific output stream and channel location. When the control bit is high, the corresponding output channel is in the high impedance state, whereas when the control bit is low, the corresponding output channel has active output data.

#### 15.1 BCSTo Control Stream

When the STio0-31 streams are in the 8 Mb/s mode, the STio0\_Ch0 control bit of the BCSTo stream is advanced by thirty-six C32/64o 32.768 Mb/s clock cycles from the backplane frame boundary. See Figure 6, "Backplane Control (BCSTo) Timing when the STio data rate is 8 Mb/s" on page 26 for the BCSTo control bit pattern.

When the STio0-15 streams are in the 16 Mb/s mode, the STio0\_Ch0 control bit of the BCSTo is advanced by thirty-six C32/64o 32.768 Mb/s clock cycles from the backplane frame boundary. See Figure 7, "Backplane Control (BCSTo) Timing when the STio data rate is 16 Mb/s" on page 27 for the BCSTo control bit pattern.



Figure 6 - Backplane Control (BCSTo) Timing when the STio data rate is 8 Mb/s



Figure 7 - Backplane Control (BCSTo) Timing when the STio data rate is 16 Mb/s

#### 15.2 LCSTo Control Stream

The LCSTo control bits are partitioned into 128 time slots, Slots 0 to 127. Each slot has 32 bits. Dummy bits represent logic levels which should be ignored by the users. The first control bit in Slot 0 is advanced by thirty-two C32/640 32.768 Mb/s clock cycles from the frame boundary. See Figure 8, "Local Control (LCSTo) Timing when STo0-18 are operated at 8 Mb/s" on page 28 for the partition of the time slots and the frame alignment details.

When the STo0-18 streams are operated at 8 Mb/s, Slots 0 to 127 are used to represent the control bits for Channels 0 to 127 of the 8 Mb/s streams. STo19-17 are driven low in this configuration. See Figure 8, "Local Control (LCSTo) Timing when STo0-18 are operated at 8 Mb/s" on page 28 for the LCSTo control bit pattern.

When the STo0-18 streams are operated at 4 Mb/s, Slots 2N (where N = 0 to 63) have the control bit pattern but Slots 2N+1 have dummy bits which should be ignored by the user. STo19-17 are driven low in this configuration. See Figure 9, "Local Control (LCSTo) Timing when STo0-18 are operated at 4 Mb/s" on page 29 for the LCSTo control bit pattern.

When the STo0-27 streams are operated at 2 Mb/s, Slots 4N (where N = 0 to 31) have the control bit pattern, but Slots 4N+1, 4N+2 and 4N+3 have dummy bits. See Figure 10, "Local Control (LCSTo) Timing when all STo0-27 are operated at 2 Mb/s" on page 30 for the LCSTo control bit pattern.

When the STo streams are programmed with various data rates as described in Table 6 on page 20, the available control bit positions in every time slot is associated with the corresponding output stream channels which operate at various data rates. Figure 11, "Example of Local Control (LCSTo) Timing when the Local Streams have Different Data Rates" on page 31 gives an example when STo0-3, 4-7, 8-11, 12-15, 16-27 are programmed to operate at 8 Mb/s, 4 Mb/s, 8 Mb/s, 4 Mb/s and 2 Mb/s respectively.





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### 16.0 DPLL

The Digital Phase Locked Loop (DPLL) accepts selectable 1.544 MHz, 2.048 MHz, or 8 kHz input reference signals. It accepts reference inputs from independent sources and provides bit-error-free reference switching. The DPLL meets phase slope and MTIE requirements defined by the Telcordia GR-1244-CORE standard.

The DPLL also provides the timing for the rest of the MT90866 Digital Switch, generating several network clocks with the appropriate quality. Clocks are synchronized to one of two input reference clocks and meet the requirements of the H.110 clock specification.

The master clock (CLK80M) for the DPLL is provided by the Analog Phase Locked Loop (APLL) from the MT90866 master clock input pin C20i. Since the APLL output is "locked" to the input, the accuracy of CLK80M clock is equal to the accuracy of C20i.

#### 16.1 MT90866 Modes of Operation

The DPLL, and consequently the MT90866, can, as required by the H.110 standard, operate in three different modes: Primary Master, Secondary Master and Slave. See Figure 12, "Typical Timing Control Configuration" on page 32.

To configure the DPLL, there are two Operation Mode registers: DOM1 and DOM2. See Table 20 on page 55 and Table 21, "DPLL Operation Mode (DOM2) Register Bits" on page 58 for the contents of these registers.

In all modes the MT90866 monitors both the "A Clocks" (C8\_A\_io and FRAME\_A\_io) and the "B Clocks" (C8\_B\_io and FRAME\_B\_io). The Fail\_A and the Fail\_B signals indicate the quality of the "A Clocks" and "B Clocks" respectively.



Figure 12 - Typical Timing Control Configuration

#### 16.1.1 Primary Master Mode

In the Primary Master Mode, the MT90866 drives the "A Clocks" (C8\_A\_io and FRAME\_A\_io), by locking to the primary reference (PRI\_REF). The PRI\_REF can be provided by one of the locally derived network reference sources (LREF0-7), or the CTREF1 input or the CTREF2 input. In this mode the MT90866 has the ability to monitor the primary reference. If the primary reference becomes unreliable, the device continues driving "A Clocks" in stable Holdover Mode until it makes a **Stratum 4 Enhanced** compatible switch to the secondary reference (SEC\_REF) for its network timing. The secondary reference can be provided by one of the local network references (LREF0-7), the CTREF1 or the CTREF2.

If the primary reference comes back or recovers, the MT90866 makes a **Stratum 4 Enhanced** compatible switch back to the original primary reference and the system returns to normal operation state.

If necessary, the MT90866 can be prevented from switching back to the original primary reference by programming the RPS bit in DOM1 register to give preference to the secondary reference.

While in the Primary Master mode, the MT90866 attenuates jitter and wander above 1.52 Hz from the selected input reference clock and generates all output clocks according to the DPLL jitter transfer function diagram on Figure 17, "DPLL Jitter Transfer Function Diagram - wide range of frequencies" on page 42 and Figure 18, "Detailed DPLL Jitter Transfer Function Diagram" on page 43.

For the Primary Master mode selection, see Table 22, "MT90866 Mode Selection - By Programming DOM1 and DOM2 Registers" on page 60.

#### 16.1.2 Secondary Master Mode

In the Secondary Master Mode, the MT90866 drives the "B Clocks" (C8\_B\_io and FRAME\_B\_io), by locking to the "A Clocks". As required by the H.110 standard, the "B Clocks" are edge-synchronous with the "A Clocks", as long as jitter on the "A Clocks" meets Telcordia GR-1244-CORE specifications.

If the "A Clocks" become unreliable, system software is notified and the MT90866 continues driving the "B Clocks" in stable Holdover Mode until it makes a **Stratum 4 Enhanced** compatible switch to the secondary reference (SEC\_REF) for its network timing. The secondary reference can be the local network reference (LREF0-7), the CTREF1 or the CTREF2. If the "A Clocks" can not recover, the designated secondary master can be promoted to primary master by system software. This promotion will cause the "B Clocks" to assume the role of the "A Clocks".

For the Secondary Master mode selection, see Table 22, "MT90866 Mode Selection - By Programming DOM1 and DOM2 Registers" on page 60.

#### 16.1.3 Slave Mode

In the Slave Mode, the MT90866 is phase locked to the "A Clocks". If the "A Clocks" become unreliable, the device goes to stable Holdover Mode until it makes a **Stratum 4 Enhanced** compatible switch to the "B Clocks". The MT90866 will perform all required functionality as long as the "A Clocks" and the "B Clocks" conform to the Telcordia GR-1244-CORE jitter specifications.

In addition, the device can be used to generate a CT reference (CT\_REF1 or CT\_REF2) from its network references, LREF0-7.

While the device is in Slave Mode and the "A Clocks" or the "B Clocks" do not recover, then the designated slave can be promoted to secondary master by system software. In that case, the network reference can be used as the secondary reference.

Table 22 on page 60 shows how to program the DOM1 and DOM2 registers to enable the Slave mode of the MT90866.

# 17.0 DPLL Functional Description



Figure 13 - DPLL Functional Block Diagram

# 17.1 Reference Select and Frequency Mode MUX Circuits

The DPLL accepts two simultaneous reference input signals and operates on their rising edges. Either the primary reference (PRI\_REF) signal or the secondary reference (SEC\_REF) signal can be selected to be the reference signal (REF) to the PLL circuit. The appropriate frequency mode input (either FREQ\_MOD\_PRI or FREQ\_MOD\_SEC) is selected to be the input of the PLL Circuit. The selection is done by the State Machine Circuit based on the current state.

The FREQ\_MOD\_PRI and the FREQ\_MOD\_SEC are 2-bit wide inputs which reflect the value in the FP1-0 and FS1-0 bits of the DOM1 register. The primary and the secondary references operate independently from each other and can have different frequencies. Switching the reference from one frequency to another does not require the device reset to be applied. Table 20 on page 55 shows input frequency selection for the primary and secondary reference respectively.

#### 17.2 PRI and SEC MUX Circuits

The DPLL has four different modes to handle reference failure. These modes are selected by the FDM0 and FDM1 bits of the DOM2 Register. If FDM1-0 is '10' then the Primary reference is always used regardless of failures. If FDM1-0 is '11' then the Secondary reference is always used regardless of failures. Otherwise the DPLL operates in one of two failure detection modes: Autodetect or Manual detection mode. When the FDM0 and FDM1 bits are set to low in the DOM2 register '00', the DPLL is in the Autodetect Mode. In this mode, the outputs from the Reference Monitor Circuits LOS\_PRI and LOS\_SEC are used by the State Machine Circuit. When the FDM0 bit is set to high and FDM1 bit is set to low '01', the DPLL is in the Manual Detection Mode and the LOS\_PRI and LOS\_SEC signals are selected from the PRI\_LOS and SEC\_LOS input pins to be used by the State Machine Circuit. See Table 21 on page 58 for selection of the Failure Detection Modes.

#### 17.3 Frame Select MUX

When the "A Clocks" or the "B Clocks" are selected as the input reference, an 8.192 MHz clock (either C8\_A\_io or C8\_B\_io) is provided to be the input reference to the PLL circuit (REF). Because the output frame pulse (CT\_FRAME) must be aligned with the selected input frame pulse, the appropriate frame pulse (either FRAME\_A\_io or FRAME\_B\_io) is selected in the Frame Select MUX circuit to be the input of the PLL circuit (FRAME).

#### 17.4 CT Clock and Frame Monitor Circuits

The CT Clock and Frame Monitor circuits check the period of the C8\_A\_io and the C8\_B\_io clocks and the FRAME\_A\_io and FRAME\_B\_io frame pulses. According to the H.110 specification, the C8 period is 122 ns with a tolerance of +/- 35 ns measured between rising edges. If C8 falls outside the range of [87 ns,157 ns], the clock is rejected and the fail signal (FAIL\_A or FAIL\_B) becomes high. The Frame pulse period is measured with respect to the C8 clock. The frame pulse period must have exactly 1024 C8 cycles. Otherwise, the fail signal (FAIL\_A or FAIL\_B) becomes high. Becomes high. When the CT BUS clock and frame pulse signals return to normal, the FAIL\_A or FAIL\_B signal returns to logic low.

#### 17.5 Reference Monitor Circuits

There are two Reference Monitor Circuits: one for the primary reference (PRI\_REF) and one for the secondary reference (SEC\_REF). These two circuits monitor the selected input reference signals and detect failures by setting up the appropriate fail outputs (FAIL\_PRI and FAIL\_SEC). These fail signals are used in the Autodetect mode as the LOS\_PRI and LOS\_SEC signals to indicate when the reference has failed. The method of generating a failure depends on the selected reference.

When the selected reference frequency is 8.192 MHz ("A Clocks" or "B Clocks"), the fail signals are passed through from the CT Clock and Frame Monitor circuit outputs FAIL\_A and FAIL\_B, and used directly as FAIL\_PRI and FAIL\_SEC, accordingly.

For all other reference frequencies (8 kHz, 1.544 MHz and 2.048 MHz), the following checks are performed:

- For all references, the "minimum 90 ns" check is done. This is required by the H.110 specifications both low level and high level of the reference must last for minimum 90 ns each.
- The "period in the specified range" check is done for all references. The length of the period of the selected input reference is checked if it is in the specified range. For the E1 (2.048 MHz clock) or the T1 (1.544 MHz clock) reference, the period of the clock can vary within the range of 1 +/- 1/4 of the defined clock period which is 488 ns for the E1 clock and 648 ns for T1 clock. For the 8 KHz reference, the variation is from 1 +/- 1/32 period.
- If the selected reference is E1 or T1, "64 periods in the specified range" check is done. The selected reference is observed for a long period (64 reference clock cycles) and checked if it is within the specified range from 62 to 66 clock periods.

These reference signal verifications include a complete loss or a large frequency shift of the selected reference signal. When the reference signal returns to normal, the LOS\_PRI and LOS\_SEC signals will return to logic low.

#### 17.6 State Machine Circuit

The State Machine handles the reference selection. Depending on REF\_SEL and LOS signals (selection between FAIL\_PRI and PRI\_LOS and between FAIL\_SEC and SEC\_LOS), the state machine selects PRI\_REF or SEC\_REF as the current input reference and dictates the PLL Circuit mode: Normal or Holdover Mode. In the Normal Mode, the DPLL output clocks are locked to the selected input reference (PRI\_REF or SEC\_REF). In the Holdover Mode, the DPLL clocks retain the phase and frequency values they had 32 to 64 ms prior to moving from the Normal to the Holdover Mode. When going from the Holdover to the Normal Mode, the State Machine activates the MTIE circuit and goes through the states MTIE PRI or MTIE SEC to prevent a phase shift of the output clocks during the DPLL reference switch (from PRI\_REF to SEC\_REF and vice versa). The state diagram is given in Figure 14, "State Machine Diagram" on page 36.



Figure 14 - State Machine Diagram

# 18.0 Phase Locked Loop (PLL) Circuit

As shown in Figure 15, "Block Diagram of the PLL Module" on page 37, the PLL module consists of a Skew Control, Maximum Time Interval Error (MTIE), Phase Detector, Phase Offset Adder, Phase Slope Limiter, Loop Filter, Digitally Controlled Oscillator (DCO), Divider and Frequency Select MUX modules.


Figure 15 - Block Diagram of the PLL Module

#### 18.1 Skew Control

The circuit delays a selected reference input with a tapped delay line with seven taps - see Figure 16, "Skew Control Circuit Diagram" on page 37. The maximum delay of the per unit delay element is factored at intervals of 3.5 ns. The tap is selected by the SKEW\_CONTROL bus which is programmed by the SKC2-SKC0 bits of the DPLL Output Adjustment (DPOA) register. The skew of this input will result in a static phase offset which varies from 0 to 7 steps of the maximum delay per unit delay element, between the input and the outputs of the DPLL.





#### 18.2 Maximum Time Interval Error (MTIE)

The MTIE circuit prevents any significant change in the output clock phase during a reference switch. Because the input references can have any relationship between their phases and the output follows the selected input reference, any switch from one reference to another could cause a large phase jump in the output clock if such a circuit did not exist. This large phase jump could cause significant data loss. The MTIE circuit keeps the phase difference between the output clock of the DPLL and the input reference the same as if the reference switch had not taken place.

The MTIE circuit has two modes:

 Measuring mode - the circuit measures the phase difference between the new reference from the Skew Control circuit and the feedback signal (FEEDBACK) from the Frequency Select MUX circuit. This mode is active during the movement of the DPLL from the Holdover to the Normal Mode, and is set by the MTIE\_START signal of the State machine module. The measured value is stored into a counter and used in the Delay mode. When the measurement process is done, the State Machine module is notified by generating the MTIE\_DONE signal, allowing it to go to the Normal Mode. Delay mode - after the rising edge of the new reference clock from the Skew Control circuit, the MTIE circuit
uses the measured value to generate the virtual reference pulse (REF\_VIR) to the Phase Detector circuit.
While the DPLL is in the Normal Mode, the MTIE Circuit is in Delay mode. It keeps the phase difference
between the output signals of the DPLL and selected input reference as the previous output signal would
have been if the reference switch had not taken place.

During a reference switch, the State Machine module first changes the mode of the DPLL from the Normal to the Holdover Mode. In the Holdover Mode, the DPLL no longer uses the virtual reference signal, but generates very accurate outputs using storage techniques.

Because the input reference coming from the Skew Control circuit is asynchronous to the sampling clock used in the MTIE circuit, a phase error may exist between the selected input reference signal and the output signal of the DPLL. In the worst case, the Maximum Time Interval Error (MTIE) is one period of the internally used clock cycle (65.536 MHz if the selected reference frequency is 8 kHz, 2.048 MHz and 8.192 MHz, and 49.408 MHz when the selected reference frequency is 1.544 MHz). This phase error is a function of the difference in phase between the two input reference signals during reference rearrangements. Each time a reference switch is made, the delay between the input signal and the output signal can change. The value of this delay is the accumulation of the error measured during each reference switch. After many switches from one reference to another, the delay between the selected input reference and the DPLL output clocks can become unacceptably large. The user should provide MTIE reset (set MRST bit in the DOM2 register to high) causing output clocks to align to the nearest edge of the selected input reference. It is recommended that the MTIE is reset after multiple reference switchings and the device falls back to its initial reference. The MTIE **MUST** be kept in the reset mode when the ZL50031 is operating in the slave mode.

#### 18.3 Phase Detector

The Phase Detector circuit compares the virtual reference signal from the MTIE Circuit (REF\_VIR) with the feedback signal from the Frequency Select MUX circuit (FEEDBACK) with respect to their rising edges, and provides an error signal corresponding to the phase difference between the two. This error signal is passed to the Phase Offset Adder Circuit. The Frequency Select MUX allows the proper feedback signal to be selected (e.g., 8 kHz, 1.544 MHz, 2.048 MHz or 8.192 MHz).

#### 18.4 Phase Offset Adder

The Phase Offset Adder Circuit adds the PHASE\_OFFSET word (bits POS6-POS0 of the DPLL Output Adjustment register - see Table 23 on page 61) to the error signal from the Phase Detector circuit to create the final phase error. This value is passed to the Phase Slope Limiter circuit. The PHASE\_OFFSET word can be positive or negative. Since the PLL will stabilize to a situation where the average of the sum of the phase offset word and the phase detector output is zero, a nonzero value in the input of the Phase Offset Adder circuit will result in a static phase offset between the input and output signals of the DPLL.

If the selected input reference of the DPLL is either 8 KHz or 2.048 MHz, the step size in this static phase offset is 15.2 ns. With a 7-bit 2's complement value, the static phase offset can be set between -0.96  $\mu$ s and +0.97  $\mu$ s. If the selected input reference of the DPLL is 1.544 MHz, the maximum phase offset is between -1.27  $\mu$ s and 1.29  $\mu$ s with a resolution of 20.2 ns.

Together with the Skew Control bits (SKC2-SKC0), users can program a static phase offset between -960 ns and +990 ns if the selected input reference of the DPLL is either 8 kHz or 2.048 MHz. If the selected reference is 1.544 MHz, the programmable phase offset is between -1.27  $\mu$ s and 1.30  $\mu$ s. For the programmable ranges mentioned above, the resolution is 1.9 ns per step. See Table 23 on page 61 for the content of the DPOA register.

When the selected input reference frequency of the DPLL is 8.192 MHz ("A Clocks" or "B Clocks" are selected as the reference), the Phase Offset Adder is bypassed. The output of the Phase Detector circuit is connected directly to the input of the Phase Slope Limiter circuit. When an 8.192 MHz clock (C8\_A\_io or C8\_B\_io) is used as the reference in the Secondary Master or the Slave mode, the H.110 standard requires the output clock to always follow the input reference on an edge-to-edge basis, so the static phase offset is not required.

#### 18.5 Phase Slope Limiter

The limiter receives the error signal from the Phase Offset Adder circuit and ensures that the DPLL responds to all input transient conditions with a maximum output phase slope of 7.6 ns per 125 us. Because of this slope, the MT90866 is within the maximum phase slope of 81 ns per 1.326 ms specified by the Telcordia GR-1244-CORE standard.

The frequency stability of the Holdover Mode is  $\pm 0.07$  ppm, which translates to a worst case 49 frame (125  $\mu$ s) slips in 24 hours. This is better than the Telcordia GR-1244-CORE Stratum 3 requirement of  $\pm 0.37$  ppm (255 frame slips per 24 hours).

#### 18.6 Loop Filter

The Loop Filter circuit gives frequency offset to the DCO circuit, based on the phase difference between the input and the feedback reference. It is similar to a first order low pass filter, with two positions for cut-off frequency (-3 dB attenuation) depending on the selected reference frequency, and it mainly determines the jitter transfer function of the DPLL.

In Primary Master mode when the selected input reference frequency is either 2.048 MHz, 1.544 MHz or 8 kHz, the cut-off frequency is approximately at 1.52 Hz and all the reference variations, including jitter, are attenuated according to the DPLL jitter transfer function (see Figure 17, "DPLL Jitter Transfer Function Diagram - wide range of frequencies" on page 42 and Figure 18, "Detailed DPLL Jitter Transfer Function Diagram" on page 43). The Loop Filter circuit ensures that the jitter transfer requirements in ETS 300-011 and Telecordia GR-499-CORE are met when the selected reference frequency is either 2.048 MHz, 1.544 MHz or 8 kHz.

When the selected input reference frequency is 8.192 MHz (i.e., in Secondary Master or Slave modes), the reference variations are bypassed to the output clocks. The cut-off frequency is at about 100 kHz, well beyond 500 Hz, the corner frequency of the Telcordia GR-1244-CORE input jitter tolerance curve.

The storage techniques, which enable generating very accurate output frequencies during the Holdover Mode of DPLL, are built into the Loop Filter circuit. When no jitter is presented on the selected input reference, the holdover frequency stability is 0.007 ppm.

#### 18.7 Digitally Controlled Oscillator (DCO)

The DCO circuit adds frequency offset from the Loop Filter, which represents the phase error between the input and the feedback reference, to the ideal center frequency value and generates appropriately corrected output high speed clock. The Synchronization method of the DCO is dependent on the state of the DPLL State Machine module.

In the Normal Mode, the DCO circuit provides an output signal which is frequency and phase locked to the selected input reference signal.

In the Holdover Mode, the DCO circuit is running at a frequency that is equal to the frequency which was generated by the DCO circuit when the DPLL was in the Normal Mode.

In the Freerun Mode, the DCO circuit is free running at its center frequency with an output accuracy equal to the accuracy of the device master clock (C20i).

#### 18.8 Divider

The Divider Circuit divides the DCO output frequency down to the required outputs. The following outputs are generated:

- C64 (65.536 MHz clock) used as the internal clock for the MT90866 device.
- CT\_C8 (8.192 MHz clock), C2M (2.048 MHz clock), C1M5o (1.544 MHz clock) and CT\_FRAME (8 kHz negative frame pulse) feedback reference signals to the Frequency Select MUX Circuit.

The CT\_FRAME and the CT\_C8 are required clocks. C1M5o is provided as an output clock of the MT90866.

The duty cycle of all output signals is independent of the duty cycle of the device master clock, C20i. The CT\_C8, C2M and C1M50 clocks have nominal 50% duty cycle,

The output frame pulse (CT\_FRAME) is generated in such a way that it is always aligned with the CT\_C8 clock to form the required H.110 CT Bus clock and frame pulse shape (when the CT\_FRAME is low the rising edge of the CT\_C8 defines the frame boundary). Depending on the selected input reference frequency, the CT\_FRAME is generated in the following way:

- When the input reference frequency is 8 kHz, the output frame pulse is aligned with the rising edge of the reference.
- When the reference frequency is either 2.048 MHz or 1.544 MHz, the CT\_FRAME randomly defines the output frame boundary, always keeping the described relation to the CT\_C8 clock.
- When the reference frequency is 8.192 MHz, the output frame pulse (CT\_FRAME) has to be aligned with the input frame pulse (FRAME\_A\_io or FRAME\_B\_io). Since an 8.192 MHz clock (either C8\_A\_io or C8\_B\_io) is used as the reference clock, the selected frame pulse from the Frame Select MUX is provided as the input to the Divider circuit and the CT\_FRAME is synchronized to it.

#### **18.9 Frequency Select MUX Circuit**

According to the selected input reference of the DPLL, this MUX will select the appropriate output frequency to be the feedback signal to the PLL and MTIE Circuits.

#### 18.10 Modes of Operation

The DPLL can operate in two main modes: the Normal and the Holdover Mode. Each of these modes has two states: the primary or the secondary state. The state depends on which reference is currently selected as the preferred reference the PRI\_REF or the SEC\_REF. When the DPLL is in the Holdover Mode and the HRST bit of the DOM2 register is pulsed logic high (or held high continuously), the DPLL operates in Freerun Mode.

#### 18.10.1 Normal Mode

Normal Mode is typically used when a clock source synchronized to the network is required.

In the Normal Mode, the DPLL provides timing (C64, CT\_C8, C2M and C1M50) and frame synchronization (CT\_FRAME) signals which are synchronized to one of two input references (PRI\_REF or SEC\_REF). The input reference signal may have a nominal frequency of 8 kHz, 1.544 MHz, 2.048 MHz or 8.192 MHz.

From a device reset condition or after reference switch, the DPLL can take up to 50 seconds to phase lock the output signals to the selected input reference signal.

#### 18.10.2 Holdover Mode

Holdover Mode is typically used for short durations while network synchronization is temporarily disrupted.

If the FDM1-0 bits are programmed to '01' in the DOM2 register and the PRI\_LOS and SEC\_LOS pins are high, the DPLL is in the Holdover Mode. The DPLL can also be in the Holdover Mode if the FDM1-0 bits are programmed to '00' and the SLS and PLS bit are observed as '11' in the DPLL House Keeping Register (DHKR).

In the Holdover Mode, the DPLL provides timing and synchronization signals which are based on storage techniques and are not locked to an external reference signal. The storage value is determined while the device is in Normal Mode and locked to an external reference signal. When the DPLL is in the Normal Mode and locks to the input reference signal, a numerical value corresponding to the DPLL output reference frequency is stored alternately in two memory locations every 32 ms. When the device is switched into the Holdover Mode, the value in memory from between 32 ms and 64 ms is used to set the output frequency of the device.

The frequency stability of the Holdover Mode is  $\pm 0.07$  ppm, which translates to a worst case 49 frame (125  $\mu$ s) slips in 24 hours.

Two factors affect the frequency stability of the Holdover Mode. The first factor is the drift on the frequency of the master clock (C20i) while in the Holdover Mode. Drift on the master clock directly affects the Holdover Mode stability. Note that the absolute master clock stability does not affect the Holdover Frequency stability, only the change in C20i stability while in Holdover. For example, a  $\pm$ 32 ppm master clock may have a temperature coefficient of  $\pm$ 0.1 ppm/ °C. So a 10 degree change in temperature, while the DPLL is in the Holdover Mode may result in an additional offset (over the  $\pm$ 0.07 ppm) in frequency stability of  $\pm$ 1 ppm, which is much greater than the  $\pm$ 0.07 ppm of the DPLL. The second factor affecting Holdover frequency stability is large jitter on the reference input prior to the mode switch.

#### 18.10.3 Freerun Mode

When the DPLL is in the Holdover Mode and the HRST bit of the DOM2 register is pulsed logic high (or held high continuously), the device is in Freerun Mode.

In Freerun Mode, the DPLL provides timing and synchronization signals which are based on the frequency of the master clock (C20i) only, and are not synchronized to the reference input signals. The frequency of the output signals is an ideal frequency with the freerun accuracy of -0.03 ppm plus the accuracy of the master clock (i.e., CT\_C8 has frequency of 8.192 MHz +/- C20i\_accuracy - 0.03 ppm).

Freerun Mode is typically used when a master clock source is required, or immediately following system power-up before network synchronization is achieved.

#### **19.0 Measures of Performance**

The following are some the DPLL performance indicators and their corresponding definitions.

#### **19.1** Intrinsic Output Jitter

Intrinsic jitter is the jitter produced by the synchronizing circuit and is measured at its output. It is measured by applying a reference signal with no jitter to the input of the device, and measuring its output jitter. Intrinsic jitter may also be measured when the device is in a non-synchronizing mode, such as freerun or holdover, by measuring the output jitter of the device. Intrinsic jitter is usually measured with various band-limiting filters depending on the applicable standards.

#### **19.2** Jitter Tolerance

Jitter tolerance is a measure of the ability of a PLL to operate properly without cycle slips (i.e., remain in lock and regain lock in the presence of large jitter magnitudes at various jitter frequencies) when jitter is applied to its reference. The applied jitter magnitude and the jitter frequency depends on the applicable standards. The input jitter tolerance of the DPLL depends on the selected reference frequency and can not exceed:  $\pm$  15 U.I. for E1 or T1 references, and  $\pm$  1 U.I. for 8 kHz references.

#### 19.3 Jitter Transfer

Jitter transfer or jitter attenuation refers to the magnitude of jitter at the output of a device for a given amount of jitter at the input of the device. Input jitter is applied at various amplitudes and frequencies, and output jitter is measured with various filters depending on the applicable standards.

In slave and secondary master mode the H.110 standard requires the "B Clocks" to be edge-synchronous with the "A Clocks", as long as jitter on the "A Clocks" meets Telcordia GR-1244-CORE specifications. Therefore in these two modes no jitter attenuation is performed

In primary master mode the jitter attenuation of the DPLL is determined by the internal 1.52 Hz low pass Loop Filter and the Phase Slope Limiter. Figure 17, "DPLL Jitter Transfer Function Diagram - wide range of frequencies" on page 42 shows the DPLL jitter transfer function diagram in a wide range of frequencies, while Figure 18, "Detailed DPLL Jitter Transfer Function Diagram" on page 43 is the portion of the diagram from Figure 17 around 0 dB of the jitter transfer amplitude. At this point it is possible to see that when operating in primary master mode the DPLL is a second order, type 2 PLL. The jitter transfer function can be described as a low pass filter to 1.52 Hz, -20 dB/decade, with peaking less then 0.5 dB.

All outputs are derived from the same signal, therefore these diagrams apply to all outputs. Since 1U.I. at 1.544 MHz (648  $ns_{PP}$ ) is not equal to 1 U.I. at 2.048 MHz (488  $ns_{PP}$ ). a transfer value using different input and output frequencies must be calculated in common units (e.g., seconds) as shown in the following example:

What is the T1 and E1 output jitter when the T1 input jitter is 20 U.I. (T1 U.I. Units) and the T1 to T1 jitter attenuation is 18 dB, for a given jittering frequency?

$$\begin{aligned} &\begin{pmatrix} -\frac{A}{20} \\ OutputT1 &= InputT1 \times 10 \\ OutputT1 &= 20 \times 10 \\ OutputE1 &= OutputT1 \times \frac{(1UIT1)}{(1UIE1)} \\ OutputE1 &= OutputT1 \times \frac{(644ns)}{(488ns)} = 3.3 UI(T1) \end{aligned}$$

Using the method mentioned above, the jitter attenuation can be calculated for all combinations of inputs and outputs.

)

Because intrinsic jitter is always present, the jitter attenuation will appear to be lower for small input jitter signals than for large ones. Consequently, accurate jitter transfer function measurements are usually made with large input jitter signals (e.g., 75% of the specified maximum jitter tolerance).



Figure 17 - DPLL Jitter Transfer Function Diagram - wide range of frequencies



Figure 18 - Detailed DPLL Jitter Transfer Function Diagram

#### **19.4 Frequency Accuracy**

Frequency accuracy is defined as the absolute tolerance of an output clock signal when the DPLL is not locked to an external reference, but is operating in the Freerun Mode. Because the output of the DCO Circuit has only discrete values, the output frequency of the DPLL has the limited accuracy of 0.03 ppm based upon the design implementation. In addition, the master clock (C20i) accuracy also directly affects the freerun accuracy. The freerun accuracy is then, 0.03 ppm plus the master clock accuracy.

#### **19.5 Holdover Frequency Stability**

Holdover frequency stability is defined as the maximum fractional frequency offset of an output clock signal when it is operating using a stored frequency value. For the DPLL, the stored value is determined while the device is in Normal Mode and locked to an external reference signal. As a result, when the DPLL is in the Normal Mode, the stability of the master clock (C20i) does not affect the holdover frequency stability because the DPLL will compensate for master clock changes while in Normal Mode. However, when the DPLL is in the Holdover Mode, the stability of the master clock does affect the Holdover frequency stability. The holdover frequency stability is 0.07 ppm assuming that the C20i frequency is held constant.

#### 19.6 Locking Range

The locking range is the input frequency range over which the DPLL must be able to pull into synchronization and to maintain the synchronization. The locking range is defined by the Loop Filter Circuit and is equal to +/- 298 ppm.

Note that the locking range is related to the master clock (C20i). If the master clock is shifted by -100 ppm, the whole locking range also shifts -100 ppm downwards to be: -398 ppm to 198 ppm.

#### 19.7 Phase Slope

The phase slope or the phase alignment speed is the rate at which a given signal changes phase with respect to an ideal signal. The given signal is typically the output signal. The ideal signal is of constant frequency and is nominally equal to the value of the final output signal or final input signal. Many telecom standards like Telcordia GR-1244-CORE state that the phase slope may not exceed a certain value, usually 81 ns/1.327 ms (61 ppm). This can be achieved by limiting the phase detector output to 61 ppm or less.

In the DPLL when operating in primary master mode the Phase Slope Limiter Circuit achieves the maximum phase slope to be: 56 ppm or 7.0 ns/125 us. When operating in secondary master or slave mode the output edges follow the input edges in accordance with the H.110 standard.

#### **19.8 Maximum Time Interval Error (MTIE)**

MTIE is the maximum peak to peak delay between a given timing signal and an ideal timing signal within a particular observation period.

For the DPLL, the maximum time interval error is less than 21 ns per reference switch.

#### 19.9 Phase Lock Time

The Phase Lock Time is the time it takes the PLL to phase lock to the input signal. Phase lock occurs when the input and the output signals are not changing in phase with respect to each other (not including jitter).

Lock time is very difficult to determine because it is affected by many factors which include:

i) initial input to output phase difference

- ii) initial input to output frequency difference
- iii) PLL loop filter
- iv) PLL limiter

Although a short phase lock time is desirable, it is not always possible to achieve due to other PLL requirements. For instance, better jitter transfer performance is achieved with a lower frequency loop filter which increases lock time, but better (smaller) phase slope performance (limiter) results in longer lock times. The DPLL loop filter and limiter were optimized to meet the Telcordia GR-499-CORE jitter transfer and Telcordia GR-1244-CORE phase alignment speed requirements. Consequently, phase lock time, which is not a standards requirement, is less than 50 seconds.

#### 20.0 Initialization of the MT90866

During power up, the  $\overline{\text{TRST}}$  pin should be pulled low to ensure that the MT90866 is in the functional mode. An external pull-down resistor is required on this pin so that the MT90866 will not enter the JTAG test mode during power up.

After power up, the contents of the connection memory can be in any state. The ODE pin should be held low after power up to keep all serial outputs in a high impedance state until the microprocessor has initialized the switching matrix. This procedure prevents two serial outputs from driving the same stream simultaneously. To ensure proper operation, a delay of 600  $\mu$ s must be applied before the first microprocessor access is performed after the RESET pin is set high

During the microprocessor initialization routine, the microprocessor should program the desired active paths through the switch. The memory block programming feature can also be used to quickly initialize the backplane and local connection memories.

When this process is completed, the microprocessor controlling the MT90866 can bring the ODE pin high to relinquish the high impedance state control.

#### 21.0 JTAG Support

The MT90866 JTAG interface conforms to the Boundary-Scan IEEE1149.1 standard. The operation of the boundary-scan circuitry is controlled by an external Test Access Port (TAP) Controller.

#### 21.1 Test Access Port (TAP)

The Test Access Port (TAP) accesses the MT90866 test functions. It consists of three input pins and one output pin as follows:

- **Test Clock Input (TCK)** TCK provides the clock for the test logic. The TCK does not interfere with any on-chip clock and thus remains independent in the functional mode. The TCK permits shifting of test data into or out of the Boundary-Scan register cells concurrently with the operation of the device and without interfering with the on-chip logic.
- **Test Mode Select Input (TMS)** The TAP Controller uses the logic signals received at the TMS input to control test operations. The TMS signals are sampled at the rising edge of the TCK pulse. This pin is internally pulled to Vdd when it is not driven from an external source.
- **Test Data Input (TDI)** Serial input data applied to this port is fed either into the instruction register or into a test data register, depending on the sequence previously applied to the TMS input. Both registers are described in a subsequent section. The received input data is sampled at the rising edge of TCK pulses. This pin is internally pulled to Vdd when it is not driven from an external source.
- **Test Data Output (TDO)** Depending on the sequence previously applied to the TMS input, the contents of either the instruction register or data register are serially shifted out towards the TDO. The data out of the TDO is clocked on the falling edge of the TCK pulses. When no data is shifted through the boundary scan cells, the TDO driver is set to a high impedance state.
- **Test Reset (TRST)** Resets the JTAG scan structure. This pin is internally pulled to Vdd when it is not driven from an external source.

#### 21.2 Instruction Register

The MT90866 uses the public instructions defined in the IEEE 1149.1 standard. The JTAG Interface contains a four-bit instruction register. Instructions are serially loaded into the instruction register from the TDI when the TAP Controller is in its shifted-IR state. These instructions are subsequently decoded to achieve two basic functions: to select the test data register that may operate while the instruction is current and to define the serial test data register path that is used to shift data between TDI and TDO during data register scanning.

#### 21.3 Test Data Register

As specified in IEEE 1149.1, the MT90866 JTAG Interface contains three test data registers:

- **The Boundary-Scan Register** The Boundary-Scan register consists of a series of Boundary-Scan cells arranged to form a scan path around the boundary of the MT90866 core logic.
- **The Bypass Register** The Bypass register is a single stage shift register that provides a one-bit path from TDI to its TDO.

The Device Identification Register - The JTAG device ID for the MT90866 is 0086614B<sub>H</sub>.

Version<31:28>:	0000
Part No. <27:12>:	0000 1000 0110 0110
Manufacturer ID<11:1>:	0001 0100 101
LSB<0>:	1

#### 21.4 BSDL

A BSDL (Boundary Scan Description Language) file is available from Zarlink Semiconductor to aid in the use of the IEEE 1149 test interface.

## 22.0 Register Descriptions

Re	ad/W	rite Addre	ess: 00	00 <sub>H</sub>											
Re	eset	Value: (	0000 <sub>F</sub>	ł											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	0	STS3	STS2	2 STS1	STS0	PRST	CBEBB	SBERB	CBERL	SBERL	0	MBP	MS2	MS1	MS0
								•							
Bit		Name	•					Des	scription						
15-1	4	Unuse	d	Reserve	ed.										
13-1	2	STS3-					and Cloc ut frame pu							ed to s	select
					STS3	STS2	ST_FPo1 P	ulse Widt	h ST_	CKo1 Fre	þ				
					0	0	244			096 MHz					
					0	1	122	ns ns		192 MHz .384 MHz		_			
						v	01								
11-10	0	STS1-					and Cloc ut frame pu							ed to s	select
					STS1	STS0	ST_FPo0 P	ulse Widt	h ST_	CKo0 Fre	þ	_			
					0	0	244	ns	4.	096 MHz					
					0	1	122	ns	8.	192 MHz		_			
					1	0	61	ns	16	.384 MHz	:				
9		PRST	r i	PRBS I	Reset: \	When h	igh, the PF	RBS trans	mitter ou	tput will I	be ir	itialize	d.		
8		CBER					Rate Clear error count								
7		SBER			ne bit e		Error Rate e test. The								gister
6		CBER					<b>Clear:</b> A lo ne BER reç	-		n of this I	oit w	ill rese	t the lo	cal inte	ernal
5		SBER					Rate Test: test result		•					ocal bit	error
4		Unuse	d	Reserv	ed. In fu	unction	al mode, th	is bit MU	ST be lov	N.					
3		MBP		progran	nming fe	eature i	amming: \ s ready for nd local co	the prog	ramming	of bit 13	to b	it 15 o	f the ba	ackplar	

#### Table 9 - Control Register (CR) Bits

Bit	Name					Description
2-0	MS2-0	Memory s memories		Bits: The	se three b	its are used to select different connection and d
			MS2	MS11	MS0	Memory Selection
			0	0	0	Local Connection Memory Low Read/Write
			0	0	1	Local Connection Memory High Read/Write
			0	1	0	Backplane Connection Memory Read/Write
			0	1	1	Local Data Memory Read
			1	0	0	Backplane Data Memory Read



Rea	d/Write	Address:	0001 <sub>H</sub>												
Res	set Val	ue: 000	00 <sub>H</sub>												
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	BMS	LG41	LG40	LG32	LG31	LG30	LG22	LG21	LG20	LG12	LG11	LG10	LG02	LG01	LG00

Bit	Name					Description			
15	Unused	Reserve	d.						
14	BMS	Backplar	ne Mod	e Sele	ct: This bit re	fers to the different	mode	for the backplan	e interface.
			Γ	BMS	-	ode Usable Streams	;		
				0	8 Mb/s	STio0 - 31			
				1	16 Mb/s	STio0 - 15	]		
13-12	LG41-LG40	4(STi16- STo19-2	27 and 7 are di	STo16 riven lo	6-27) of the lo ow.	e two bits refer to d cal interface. When	opera	•	• •
		-			Switching Mo				
		_	0	0	8 Mb/s	STi16-18, STo1			
		_	0	1	4 Mb/s	STi16-18, STo1			
			1	0	2 Mb/s	STi16-27, STo1	6-27		
11-9	LG32-LG30		•		d STo12-15) c	e three bits refer to f the local interface Switching Mode	<b>)</b> .	ent switching mo	ode for
			LG32	_		8 Mb/s			4
				0	-			12-15, STo12-15	4
			0	0	-	4 Mb/s		12-15, STo12-15	4
			0	1	•	2 Mb/s		12-15, STo12-15	4
			0	1	-	4-bit wide subrate		12-15, STo12-15	4
			1	0	0	2-bit wide subrate	STI	12-15, STo12-15	

#### Table 10 - Device Mode Selection (DMS) Register Bits

Bit	Name					Description		
8-6	LG22-LG20		•			se three bits refer to ne local interface.	different switching mode t	for
			LG22	LG21	LG20	Switching Mode	Usable Streams	
			0	0	0	8 Mb/s	STi8-11, STo8-11	
			0	0	1	4 Mb/s	STi8-11, STo8-11	
			0	1	0	2 Mb/s	STi8-11, STo8-11	
			0	1	1	4-bit wide subrate	STi8-11, STo8-11	
			1	0	0	2-bit wide subrate	STi8-11, STo8-11	
		group i	LG12	LG11	LG10	local interface.	Usable Streams	
		<b>U</b> 1	•		,			
			0	0	0	8 Mb/s	STi4-7, STo4-7	
			0	0	1	4 Mb/s	STi4-7, STo4-7	
			0	1	0	2 Mb/s	STi4-7, STo4-7	
			U		U		3114-7, 3104-7	
			0	1	1	A-bit wido subrato		
			0	1	1	4-bit wide subrate	STi4-7, STo4-7	
			0 1	1 0	1 0	4-bit wide subrate 2-bit wide subrate		
2-0	LG02-LG00		1 Broup 0 N	0 Node Sel	0 ect: Thes	2-bit wide subrate	STi4-7, STo4-7	for
2-0	LG02-LG00		1 Broup 0 N	0 Node Sel	0 ect: Thes	2-bit wide subrate se three bits refer to	STi4-7, STo4-7 STi4-7, STo4-7	for
2-0	LG02-LG00		1 Group 0 N (STi0-3 a	0 Node Sel	0 ect: Thes -3) of the	2-bit wide subrate se three bits refer to local interface.	STi4-7, STo4-7 STi4-7, STo4-7 different switching modes	for
2-0	LG02-LG00		1 Group 0 N (STi0-3 a LG02	0 Node Sela and STo0 LG01	0 ect: Thes -3) of the LG00	2-bit wide subrate se three bits refer to local interface. Switching Mode	STi4-7, STo4-7 STi4-7, STo4-7 different switching modes Usable Streams	for
2-0	LG02-LG00		1 Group 0 M (STi0-3 a LG02 0	0 Node Seland STo0 LG01 0	0 ect: Thes -3) of the LG00 0	2-bit wide subrate se three bits refer to local interface. Switching Mode 8 Mb/s	STi4-7, STo4-7 STi4-7, STo4-7 different switching modes Usable Streams STi0-3, STo0-3	for
2-0	LG02-LG00		1 Group 0 N (STi0-3 a LG02 0 0	0 Node Sel and STo0 LG01 0 0	0 ect: Thes -3) of the LG00 0 1	2-bit wide subrate se three bits refer to local interface. Switching Mode 8 Mb/s 4 Mb/s	STi4-7, STo4-7 STi4-7, STo4-7 different switching modes Usable Streams STi0-3, STo0-3 STi0-3, STo0-3	for

Table 10 - Device Mode Selection (DMS) Register Bits (continued)

		te Add ue: 00		0002 <sub>H</sub>	ł										
	15 0	14 0	13 0	12 0	11 0	10 0	9 0	8 BBPD2	7 BBPD1	6 BBPD0	5 LBPD2	4 LBPD1	3 LBPD0	2 BPE	1 0
	1		 												
Bit		Namo	e							Desc	cription				
15-9	U	Inuse	ed	Res	erve	<b>d.</b> In	fund	ctional m	ode, thes	e bits Ml	JST be lo	ow.			
8-6	B	3PD2	2-0	the l is ac the o	backı ctivat conte	olane ed. A ents c	e cor fter of th	nnection the MBP e bits BB	memory l bit in the PD2 - 0 a	olock whe control r are loade	enever th egister is d into bit	e Memo set to hi s 15 - 13	ry Block I gh and th of the b	Progra ne BPE ackpla	e loaded into imming feature E is set to high, ine connection o be zero.
5-3	LE	3PD2	2-0	loca activ the o men	l con /ated conte nory	necti I. Afte ents c Iow. I	on r er th of th Bits	nemory I e MBP b e bits LB 12 - 0 of	ow when it in the c PD2 - 0 a the local	ever the control reg are loade	Memory gister is s d into bits on memo	Block Pr set to hig s 15 - 13 ory low a	ogrammi h and the of the lo	ing fea e BPE cal co	is set to high,
2		BPE		Prog defin fram BPE MBF bits Whe	gram ned in es to bit r car in the eneve	ming n the o com eturn n be s e BPI er the the u	fun san nple is to set to M re e mio iser	ction. The ne write of te the blo low to in p low to a gister mus croproces must ma	e BPE, B operation ock progra dicate th abort the ust not be ssor write intain the	BPD2-0 : . Once th amming. at the op programi e change es BPE to	and LBP e BPE bi After the eration is ming ope d for two be high gical valu	D2-0 in t it is set to block pr completeration. V frames to to start t	he BPM o high, M ogrammi ce. When When BP o ensure he block	registe T9086 ing ha BPE i BPE i E is hi prope	Memory Block er have to be 66 requires two s finished, the s high, BPE or gh, the other er operation. amming e BPM register
1-0	U	Inuse	ed					-	-	e bits Ml		OW.			
	1														

Table 11 - Block Programming Mode (BPM) Register Bits

Reau/N	Write A	Address	es: 00	004 <sub>H</sub> fo	r LIDR	0 regis	ter,	0005 <sub>H</sub>	for LI	DR1 reg	gister,				
			00	006 <sub>H</sub> fo	r LIDR	2 regis	ter,	0007 <sub>H</sub>	for LIE	DR3 reg	gister,				
			00	008 <sub>H</sub> fo	r LIDR	4 regis	ter,	0009 <sub>H</sub>	for LIE	DR5 reg	gister,				
			00	00A <sub>H</sub> fo	or LIDR	6 regis	ster,	000B <sub>H</sub>	<sub>l</sub> for LII	DR7 reg	gister,				
			00	00C <sub>H</sub> fo	or LIDF	8 regis	ster,	000D <sub>H</sub>	<sub>1</sub> for Lll	DR9 re	gister,				
Reset	Value	0000 <sub>H</sub>													
		10	10		10			_		_					
15 LIDR0 0	-	13 LID23	12 LID22	11 LID21	10	9	8	7 LID12	6 LID11	5	4	3	2 LID02	1	0
	LIDZ	LID23	LIDZZ	LIDZI	LID20	LID14	LID13	LIDTZ	LIDTI	LID10	LID04	LID03	LIDUZ	LID01	LID00
LIDR1 0	LID54	LID53	LID52	LID51	LID50	LID44	LID43	LID42	LID41	LID40	LID34	LID33	LID32	LID31	LID30
_							•								
LIDR2 0	LID84	LID83	LID82	LID81	LID80	LID74	LID73	LID72	LID71	LID70	LID64	LID63	LID62	LID61	LID60
LIDR3 0	LID11	4 LID113	LID112	LID111	LID110	LID104	LID103	LID102	LID101	LID100	LID94	LID93	LID92	LID91	LID90
LIDR4 0	LID14	4 LID143	LID142	LID141	LID140	LID134	LID133	LID132	LID131	LID130	LID124	LID123	LID122	LID121	LID120
LIDR5 0	LID17	4 LID173	LID172	LID171	LID170	LID164	LID163	LID162	LID161	LID160	LID154	LID153	LID152	LID151	LID150
				l		l			l	l		l	l	l	l
LIDR6 0	LID20	4 LID203	LID202	LID201	LID200	LID194	LID193	LID192	LID191	LID190	LID184	LID183	LID182	LID181	LID180
LIDR7 0	LID23	4 LID233	LID232	LID231	LID230	LID224	LID223	LID222	LID221	LID220	LID214	LID213	LID212	LID211	LID210
LIDR8 0	LID26	4 LID263	LID262	LID261	LID260	LID254	LID253	LID252	LID251	LID250	LID244	LID243	LID242	LID241	LID240
LIDR9 0	0	0	0	0	0	0	0	0	0	0	LID274	LID273	LID272	LID271	LID270
L		<b>I</b>	1	1	1	1	1	1	1	1		1	1	1	1

Name	Description
LIDn4, LIDn3, LIDn2, LIDn1, LIDn0 (See Note 1)	<b>Local Input Delay Bits 4 - 0:</b> These five bits define how long the serial interface receiver takes to recognize and to store bit 0 from the STi input pins: i.e., to start a new frame. The input delay can be selected to +7.75 data rate clock periods from the frame boundary.
Note 1: n denotes an ST	ï stream number from 0 to 27.

Table 12 - Local Input Bit Delay Registers (LIDR0 to LIDR9) Bits

		Corres	ponding De	lay Bits	
Local Input Bit Delay	LIDn4	LIDn3	LIDn2	LIDn1	LIDn0
No clock period shift (Default)	0	0	0	0	0
+ 1/4 data rate clock period	0	0	0	0	1
+ 1/2 data rate clock period	0	0	0	1	0
+ 3/4 data rate clock period	0	0	0	1	1
+ 1 data rate clock period	0	0	1	0	0
+ 1 1/4 data rate clock period	0	0	1	0	1
+ 1 1/2 data rate clock period	0	0	1	1	0
+ 1 3/4 data rate clock period	0	0	1	1	1
+ 2 data rate clock period	0	1	0	0	0
		L		1	ı
+ 7 3/4 data rate clock period	1	1	1	1	1

 Table 13 - Local Input Bit Delay Programming



Figure 19 - Local Input Bit Delay Timing

Read/W	/rite Ac	dress	es:	001E	C <sub>H</sub> for B E <sub>H</sub> for B	80AR2	2 regis	ster,				.R1 reg R3 regi				
Reset v	alue:			0000	) <sub>H</sub> for a	II BOA	R reg	jisters.								
	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
BOAR0	BOA 71	BOA 70	BOA 61	BOA 60	BOA 51	BOA 50	BOA 41	BOA 40	BOA 31	BOA 30	BOA 21	BOA 20	BOA 11	BOA 10	BOA 01	BOA 00
BOAR1	BOA 151	BOA 150	BOA 141	BOA 140	BOA 131	BOA 130	BOA 121	BOA 120	BOA 111	BOA 110	BOA 101	BOA 100	BOA 91	BOA 90	BOA 81	BOA 80
BOAR2	BOA 231	BOA 230	BOA 221	BOA 220	BOA 211	BOA 210	BOA 201	BOA 200	BOA 191	BOA 190	BOA 181	BOA 180	BOA 171	BOA 170	BOA 161	BOA 160
BOAR3	BOA 311	BOA 310	BOA 301	BOA 300	BOA 291	BOA 290	BOA 281	BOA 280	BOA 271	BOA 270	BOA 271	BOA 260	BOA 251	BOA 250	BOA 241	BOA 240
	Name	)							D	escript	tion					
	an1, B( ee Not		o	fset th	a <b>ne Ou</b> at a pai itput str	rticula	r strea	am outp	out ca	n be ac	lvanc	ed. Wh	en the			
			o	fset th	at a pai	rticula	r strea nas no	am outp	out ca lignm <sub>ut</sub>	n be ac	tvanc the f	ed. Wh	en the ulse.		is zero Mb/s	
			o	fset th	at a pai itput str	rticula ream l	r strea nas no	am outp ormal a	out ca lignm <sup>ut</sup> ment	n be ac ent with C8_A or C8_E	tvanc the f	ed. Wh frame p	en the ulse.	offset	is zero Mb/s	
			o	fset th	at a pai itput str BOAn1	rticula ream f	r strea nas no	am outp ormal a Outpu Advance	out ca lignm <sup>ut</sup> ment	n be ac ent with C8_A or C8_E perio	tvanc n the f <sup>io</sup> 3_io d	ed. Wh frame p 8.192 M (bit)	en the ulse. <sup>b/s</sup>	offset 16.384 (bit	is zero	
			o	fset th	at a pai itput str BOAn1 0	rticula ream ł BOA	n strea nas no	am outp ormal a Outpu Advance 0 ns	out ca lignm ut ment s	n be ac ent with C8_A or C8_E perio	the f the f <u>io</u> <u>3_io</u> d	ed. Wh frame p 8.192 M (bit) 0	en the ulse. <sup>b/s</sup>	offset 16.384 (bit	Mb/s	

Table 14 - Backplane Output Advancement Registers (BOAR0 to BOAR3) Bit





#### MT90866

Read/W	rite A	ddres	ses:		(	0020 <sub>F</sub>	<sub>1</sub> for L	OAR	0 regi	ster,		0021	<sub>H</sub> for	LOAF	R1 reg	jister,
					(	)022 <sub>F</sub>	<sub>1</sub> for L	OAR	2 regi	ster,		0023	H for	LOAF	R3 reg	jister,
Reset va	alue:				(	0000 <sub>H</sub>	<sub>I</sub> for a	ll LO	AR re	gister	S.					
	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
LOAR0	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA
	71	70	61	60	51	50	41	40	31	30	21	20	11	10	01	00
													•			
LOAR1	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA
	151	150	141	140	131	130	121	120	111	110	101	100	91	90	81	80
LOAR2	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA
	231	230	221	220	211	210	201	200	191	190	181	180	171	170	161	160
LOAR3	0	0	0	0	0	0	0	0	LOA	LOA	LOA	LOA	LOA	LOA	LOA	LOA
									271	270	271	260	251	250	241	240
												•				

Name					Descrip	tion		
LOAn1, LOAn0 (See Note 1)	that	a partic	ular strea	ncement Bits am output can ormal alignme	be advance	ed. When the		
		LOAn1	LOAn0	Output Advancement	C8_A_io or C8_B_io period	2.048 Mb/s (bit)	4.096M b/s (bit)	8.192 Mb/s (bit)
		0	0	0 ns	0	0	0	0
		0	1	- 7.5 ns	- 1/16	- 1/64	- 1/32	- 1/16
		1	0	- 15 ns	- 1/8	- 1/32	- 1/16	- 1/8
		1	1	- 22.5 ns	- 3/16	- 3/64	- 3/32	- 3/16

#### Table 15 - Local Output Advancement Registers (LOAR0 to LOAR3) Bits



#### Figure 21 - Local Output Advancement Timing

	et Val		ress: 002 00 <sub>H</sub>	- / H											
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	LBS A4	LBS A3	LBS A2	LBS A1	LBS A0	LBC A7	LBC A6	LBC A5	LBC A4	LBC A3	LBC A2	LBC A1	LBC A0

Bit	Name	Description
15 - 13	Unused	Reserved.
12 - 8	LBSA4 - LBSA0	<b>Local BER Input Stream Address Bits:</b> These bits refer to the local input data stream which receives the BER data.
7 - 0	LBCA7 - LBCA0	<b>Local BER Input Channel Address Bits</b> : These bits refer to the local input channel which receives the BER data.

#### Table 16 - Local Bit Error Rate Input Selection (LBIS) Register Bits

		ess: 0028 <sub>H</sub> e: 0000 <sub>H</sub>													
15	14	13	12	11										0	
LBER 15	LBE 14	R LBER 13	LBER 12	LBER 11	LBER 10	LBER 9	LBER 8	LBER 7	LBER 6	LBER 5	LBER 4	LBER 3	LBER 2	LBER 1	LBER 0
Bit	:	Na	me					n							
									200	criptio					

#### Table 17 - Local Bit Error Rate Register (LBERR) Bits

		e Addre	ess: 0029 0 <sub>H</sub>	) <sub>H</sub>												
15	14	13	12	11		10	9	8	7	6	5	4	3	2	1	0
0	0	0	BBSA 4	BBS 3	A	BBSA 2	BBSA 1	BBSA 0	BBCA 7	BBCA 6	BBCA 5	BBCA 4	BBCA 3	BBCA 2	BBCA 1	BBCA 0
Bit	Name     Description															
15 - 13																
12 - 8	B	BBSA4 - BBSA0       Backplane BER Input Stream Address Bits: These bits refer to the backplane input data stream which receives the BER data.										kplane				
7 - 0	- 0 BBCA7 - BBCA0 Backplane BER Input Channel Address Bits: These bits refer to the backplane input channel which receives the BER data.															

# backplane input channel which receives the BER data. Table 18 - Backplane Bit Error Rate Input Selection (BBIS) Register Bits

	d Addres et Value:		1												
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
BBER 15	BBER 14	BBER 13	BBER 12	BBER 11	BBER 10	BBER 9	BBER 8	BBER 7	BBER 6	BBER 5	BBER 4	BBER 3	BBER 2	BBER 1	BBER 0

Bit	Name	Description
15 - 0	BBER15 -BBER0	Backplane Bit Error Rate Count Bits: These bits refer to the backplane bit
		error count. This counter stops incrementing when it reaches the value 0xFFFF.

#### Table 19 - Backplane Bit Error Rate Register (BBERR) Bits

15 14 13 12 11 10 9 8 7 6 5 4 3 2 1 0 CNEN BEN AEN RPS FS1 FS0 FP1 FP0 SS3 SS2 SS1 SS0 SP3 SP2 SP1 SP0	Read/Writ Reset Val			<sub>H</sub> for DO	M1 Reg	jister											
CNEN BEN AEN RPS FS1 FS0 FP1 FP0 SS3 SS2 SS1 SS0 SP3 SP2 SP1 SP0	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	
	CNEN	BEN	AEN	RPS	FS1	FS0	FP1	FP0	SS3	SS2	SS1	SS0	SP3	SP2	SP1	SP0	

Bit	Name			Description						
15	CNEN	NREFo Output Ena When CNEN is high		CNEN is low, NREFo output is disabled, i.e. tri-s t is enabled.	stated.					
14	BEN			n BEN is low, the "B Clocks" (C8_B_io and FRA o and FRAME_B_io behave as inputs.	ME_B_io)					
		When BEN is high, outputs.	the "B Clocks" a	re enabled - C8_B_io and FRAME_B_io behav	e as					
13	AEN	are disabled, i.e. tri	-stated - C8_A_i	n AEN is low, the "A Clocks" (C8_A_io and FRA o and FRAME_A_io behave as inputs. are enabled - C8_A_io and FRAME_A_io behav						
12	12 RPS <b>Reference Selection Bit:</b> When RPS is low, the preferred reference is the primary reference (PRI_REF). When RPS is high, the preferred reference is the secondary reference (SEC_REF).									
11 - 10	FS1 - FS0	SEC_REF Frequer frequencies for the		its: These bits are used to select different clock ence.	ζ					
		FS1	FS0	Secondary Reference						
		0	0	8 kHz						
	0 1 1.544 MHz									
		1 0 2.048 MHz								
		1	1	8.192 MHz ("A Clocks" or "B Clocks")						
				Mode (DOM1) Pegister Bite						

#### Table 20 - DPLL Operation Mode (DOM1) Register Bits

Bit	Name			Description
9 - 8	FP1 - FP0		uency Selection B the primary referen	Bits: These bits are used to select different clock
			the primary referen	
		FS1	FS0	Primary Reference
		0	0	8 kHz
		0	1	1.544 MHz
		1	0	2.048 MHz
		1	1	8.192 MHz ("A Clocks" or "B Clocks")
7 - 4	SS3 - SS0	Secondary Clo reference input.		ut Selection Bits: These bits are used to select secondary
			Secondary Clock	< Reference Input
		0000	CTR	REF1
		0001	CTR	REF2
		0010	"A Cle	locks"
		0011	"B Cle	locks"
		0100	Rese	erved
		0101	Rese	erved
		0110	Rese	erved
		0111	Rese	erved
		1000	LRE	EF0
		1001	LRE	EF1
		1010	LRE	EF2
		1011	LRE	EF3
		1100	LRE	EF4
		1101	LRE	EF5
		1110	LRE	EF6
		1111	LRE	EF7
			•	

 Table 20 - DPLL Operation Mode (DOM1) Register Bits (continued)

Bit	Name		Description	
3 - 0	SP3 - SP0	Primary Clock Refere reference input.	nce Input Selection Bits: These bits are u	sed to select primary
		SP3 - SS0	Primary Clock Reference Input	
		0000	CTREF1	
		0001	CTREF2	
		0010	"A Clocks"	
		0011	"B Clocks"	
		0100	Reserved	
		0101	Reserved	
		0110	Reserved	
		0111	Reserved	
		1000	LREF0	
		1001	LREF1	
		1010	LREF2	
		1011	LREF3	
		1100	LREF4	
		1101	LREF5	
		1110	LREF6	
		1111	LREF7	

Table 20 - DPLL Operation Mode (DOM1) Register Bits (continued)

	d/Write et Valu			02C <sub>H</sub> for	DOM2 R	egister									
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	HRST	MRST	FDM1	FDM0	BFEN	AFEN	CNIN	DIV1	DIV0	CNS2	CNS1	CNS0

Bit	Name			Description							
15 - 12	Unused	Reserved	d.								
11	HRST	circuit is reset. Wł	in function ile the D	ory Reset Bit: When HRST is low, the DPLL hold memory onal mode. When HRST is high, the hold memory circuit will be OPLL is in Holdover Mode, pulsing HRST high (or holding it high force the DPLL to the Freerun Mode.							
10	MRST	When MF with the r	RST is hi nearest e	When MRST is low, the DPLL MTIE circuit is in functional mode. igh, the MTIE circuit will be reset - the DPLL outputs will align edge of the selected reference. When the MT90866 is operating e, this bit <b>MUST</b> be set high to reset the MTIE circuit.							
9 - 8	FDM1 - FDM0	Failure Detection		ode Bits: These two bits control how to choose the Failure							
		FDM1	FDM0	Failure Detection Mode							
		0	0	Autodetect - Automatic Failure Detection by internal reference monitor circuit							
		0	1	External - Failure Detection controlled by external inputs (PRI_LOS and SEC_LOS)							
		1	0	Forced Primary - The DPLL is forced to use primary reference							
		1	1	Forced Secondary - The DPLL is forced to use secondary reference							
7	BFEN	<b>B Clocks Fail Output Enable Bit:</b> When BFEN is low, FAIL_B output is disabled, i.e., tri-stated. When BFEN is high, FAIL_B output is enabled.									
6	AFEN		A Clocks Fail Output Enable Bit: When AFEN is low, FAIL_A output is disabled, i.e., tri-stated. When AFEN is high, FAIL_A output is enabled.								
5	CNIN	CTREF2	inputs w	<b>REF2 Inputs Inverted:</b> When CNIN is high, the CTREF1 and ill be inverted, prior to entering the DPLL module. When CNIN is and CTREF2 inputs will not be inverted.							

Table 21 - DPLL Operation Mode (DOM2) Register Bits

Bit	Name				D	Description					
4 -3	DIV1 - DIV0		<b>Divider Bits:</b> These two bits define the relationship between the input re and the NREFo output.								
			DIV1	NREFo Output							
			0	0	Input refere	nce					
		_	0	1		nce/193 (8 KHz signal when input lock = 1.544 MHz)					
		_	1	0		nce/256 (8 KHz signal when input lock = 2.048 MHz)					
			1	1							
2 - 0	CNS2 - CNS0			the NRE	Fo source.	ese three bits select one of the LREF7 -					
			0			LREF0					
			0	0	1	LREF0					
			0	1	0	LREF1					
			0	1	1	LREF2 LREF3					
			1	0	0	LREF3					
			1	0	1	LREF5					
			1	1	0	LREF6					
			1	1	1	LREF7					
				<u> </u>							

Table 21 - DPLL Operation Mode (DOM2) Register Bits (continued)

	Mode	Secondary Master Mode	Slave Mode		
BEN (bit 14)	0 - Monitor "B Clocks"	1 - Drive "B Clocks"	0 - Monitor "B Clocks"		
AEN (bit 13)	1 - Drive "A Clocks"	0 - Monitor "A Clocks"	0 - Monitor "A Clocks"		
RPS (bit 12)	0 - Preferred reference is PRI_REF	0 - Preferred reference is PRI_REF	0 - Preferred reference is PRI_REF		
FS1-0 (bits 11-10) Frequency of the secondary reference	00 - 8 kHz 01 - 1.544 MHz 10 - 2.048 MHz	00 - 8 kHz 01 - 1.544 MHz 10 - 2.048 MHz	11 - 8.192 MHz Clock ("B Clocks")		
FP1-0 (bits 9-8) Frequency of the primary reference	00 - 8 kHz 01 - 1.544 MHz 10 - 2.048 MHz	<b>11</b> - 8.192 MHz Clock ("A Clocks")	11 - 8.192 MHz Clock ("A Clocks")		
SS3-0 (bits 7-4) Secondary reference selection: SP3-0 (bits 3-0) Primary reference selection:	0000 - CTREF1 0001 - CTREF2 1000 - LREF0 1001 - LREF1 1010 - LREF2 1011 - LREF3 1100 - LREF4 1101 - LREF5 1110 - LREF6 1111 - LREF6 1111 - LREF7 0000 - CTREF1 0001 - CTREF2 1000 - LREF0 1001 - LREF1 1010 - LREF3 1100 - LREF3 1100 - LREF4 1101 - LREF5 1110 - LREF6 1111 - LREF7	0000 - CTREF1 0001 - CTREF2 1000 - LREF0 1001 - LREF1 1010 - LREF2 1011 - LREF3 1100 - LREF4 1101 - LREF5 1110 - LREF6 1111 - LREF6 1111 - LREF7 XXXX - C8_A_io When bits <b>FP1-0</b> are set to <b>11</b> , C8_A_io is always used as the primary reference, regardless of the values of bits SP3-0. Output frame pulses are aligned to FRAME_A_io if primary reference is the active reference	<ul> <li>XXXX - C8_B_io</li> <li>When bits FS1-0 are set to 11, C8_B_io is always used as the secondary reference, regardless of the values of bits SS3-0. Output frame pulses are aligned to FRAME_B_io if secondary reference is the active reference</li> <li>XXXX - C8_A_io</li> <li>When bits FP1-0 are set to 11, C8_A_io is always used as the primary reference, regardless of the values of bits SP3-0. Output frame pulses are aligned to FRAME_A_io if primary reference is the active reference</li> </ul>		
MRST (bit 10)	0 - MTIE functional 1 - MTIE reset	0 - MTIE functional 1 - MTIE reset	1 - MTIE MUST be kept in the reset state in Slave mode		
FDM1, FDM0 (bits 9-8) Failure detect mode selection	00 - Autodetect Mode	00 - Autodetect Mode 01 - External Mode (Note 1)	00 - Autodetect Mode 01 - External Mode (Note 1)		
	RPS (bit 12) FS1-0 (bits 11-10) Frequency of the secondary reference FP1-0 (bits 9-8) Frequency of the primary reference SS3-0 (bits 7-4) Secondary reference selection: SP3-0 (bits 3-0) Primary reference selection: MRST (bit 10) FDM1, FDM0 (bits 9-8) Failure detect mode selection	RPS (bit 12)0 - Preferred reference is PRI_REFFS1-0 (bits 11-10) Frequency of the secondary reference00 - 8 kHz 01 - 1.544 MHz 10 - 2.048 MHzFP1-0 (bits 9-8) Frequency of the primary reference00 - 8 kHz 01 - 1.544 MHz 10 - 2.048 MHzSS3-0 (bits 7-4)000 - CTREF1 0001 - CTREF2 1000 - LREF0Secondary reference00000 - CTREF1 0001 - CTREF2 1000 - LREF0Secondary reference101 - LREF1 1010 - LREF1 1000 - LREF0Secondary reference0000 - CTREF1 0001 - CTREF2 1000 - LREF0SP3-0 (bits 3-0)0000 - CTREF1 1001 - LREF3 1110 - LREF6 1111 - LREF3 1000 - LREF0Primary reference selection:0000 - CTREF1 1001 - LREF2 1000 - LREF0 1001 - LREF5 1110 - LREF6 1111 - LREF5 1110 - LREF6 	RPS (bit 12)0 - Preferred reference is PRI_REF0 - Preferred reference is PRI_REFFS1-0 (bits 11-10)00 - 8 kHz 01 - 1.544 MHz00 - 8 kHz 01 - 1.544 MHz01 - 1.544 MHz 10 - 2.048 MHzFP1-0 (bits 9-8) Frequency of the primary reference00 - 8 kHz 01 - 1.544 MHz11 - 8.192 MHz Clock ("A Clocks")FP1-0 (bits 9-8) Frequency of the primary reference00 - 0 - 7 REF1 10 - 2.048 MHz0000 - CTREF1 0001 - CTREF2 1000 - LREF0SS3-0 (bits 7-4)0000 - CTREF1 1001 - LREF10000 - CTREF1 1000 - LREF0Secondary reference1001 - LREF2 1010 - LREF21010 - LREF2 1010 - LREF2Secondary reference101 - LREF3 1110 - LREF31011 - LREF3 1110 - LREF3SP3-0 (bits 3-0)0000 - CTREF1 1001 - CTREF2 1000 - LREF00000 - CTREF1 1010 - LREF5 1110 - LREF5 1110 - LREF6 1111 - LREF5SP3-0 (bits 3-0)0000 - CTREF1 1000 - LREF0XXXX - C8_A_io Vhen bits FP1-0 are set to 11, C8_A_io is always used as the primary reference selection:SP3-0 (bits 3-0)0000 - CTREF1 1001 - LREF6 1111 - LREF3 1100 - LREF6 1111 - LREF3 1100 - LREF4 1101 - LREF5 1110 - LREF6 1111 - LREF3 1100 - LREF4 1101 - LREF5 1110 - LREF6 1111 - LREF5 1110 - LREF6 1111 - LREF5 1110 - LREF6 		

### Table 22 - MT90866 Mode Selection - By Programming DOM1 and DOM2 Registers

	Write Ac Value: (		)02D <sub>H</sub> fo	or DPOA	Register										
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
POS 6	POS 5	POS 4	POS 3	POS 2	POS 1	POS 0	0	0	0	0	0	0	SKC2	SKC1	SKC0
Bit Name Description															
15 0		26 DC	160	<b>Phase Offset Bits:</b> These seven hits refer to the 2's complement phase word to											

15 - 9	POS6 - POS0	<b>Phase Offset Bits:</b> These seven bits refer to the 2's complement phase word to control the DPLL output phase offset. The offset varies in steps of 15 ns if the reference is 8 kHz or 2.048 MHz. The offset varies in steps of 20 ns if the reference is 1.544 MHz.
8 - 3	Unused	Reserved.
2 - 0	SKC2 - SKC0	<b>Skew Control Bits:</b> These three bits control the delay of the DPLL outputs from 0 to 7 steps in interval of maximum unit delay of 3.5 ns.

Table 23 - DPLL Output Adjustment (DPOA) Register Bits

Rea	ad/W	rite Add	lres	s: 00	02E <sub>H</sub>	for DH	IKR R	egister	-							
Res	set Va	alue: 00	00 <sub>H</sub>	I												
15	14	13	12	2	11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0		0	0	0	0	0	SLS	PLS	CKM	Limit	State 2	State 1	State 0
Bit		Name								De	escript	ion				
15 - 7	'	Unused	b	Re	serv	ed.										
6		SLS			<b>Example 2 Constant of Secondary Loss Detection Bit (Read-only bit):</b> This bit is the same as the output om the DPLL Reference Monitor FAIL_SEC.											
																-

5	PLS	<b>Primary Loss Detection Bit (Read-only bit):</b> This bit is the same as the output from the DPLL Reference Monitor FAIL_PRI.
4	СКМ	<b>DPLL Clock Monitor Bit:</b> When high, the primary output C32/64o is 65.536 MHz clock. When low, the primary output C32/64o is 32.768 MHz clock. This is the only writable bit in this register.
3	Limit	Limit (Read-only bit): Indicates that DPLL Phase Slope limiter limits input phase.

Table 24 - DPLL House Keeping (DHKR) Register Bits

2 - 0	State	<b>State:</b> These 3 bits in 14, "State Machine D		the DPLL State Machine. F 6.	Please refer to Fig
			State 2-0	State Name	
			000	NORMAL_PRI	
			001	Reserved	
			010	HOLDOVER_PRI	
			011	MTIE_PRI	
			100	NORMAL_SEC	
			101	Reserved	
			110	HOLDOVER_SEC	
			111	MTIE_SEC	

Table 24 - DPLL House Keeping (DHKR) Register Bits (continued)

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
BTM	BTM	BTM	BSAB	BSAB	BSAB	BSAB	BSAB	BCA							
2	1	0	4	3	2	1	0	B7	B6	B5	B4	B3	B2	B1	B0

15 -13							DC	scriptic					
	BTM2 - 0	Throu CT-Bu					sage Contr	ol Bits:	These t	hree bits	control	the back	plane
		<u>BTM2</u> 000		Per me	-chai mory	is the loca	and Messa le delay fro I data mem	m local ory add	interface ress of tl	; the con he switcl	hed inpu	ut channe	
		001		Per- mer	-char mory	nnel consta is the loca	ackplane C ant delay fro Il data mem	om local lory add	interface ress of tl	e; the co he switcl	ntent of hed inpu	the conne	
		010	)	Per con inpu	-chai necti ut cha	nnel variat	ackplane C le delay fro y is the bac stream. The	m backp kplane o	blane inte data mer	erface; t nory ado	he conte dress of	ent of the the switc	hed
		011		Per con inpu	r-cha necti ut cha	nnel const on memor	ant delay fr y is the bac stream. The	kplane o	data mer	nory add	dress of	the switc	hed
		100	)	Per-channel message mode; only the lower byte (bits 7 to 0) of the connection memory location will be the presented to the backplane CT-Bus output channel.									
		101		Per pre	-char sente	nnel BER p ed to the ba	battern; the ackplane C	T-Bus ou	utput cha	annel.	st patter	n will be	
		110 111			-cnar erve	•	The backpl	ane CI-	Bus is in	iput.			
			BTM2	BTM1	втмо		Source	Var. delay	Const. delay	Msg Mode	BER	I/O HiZ	
			ш 0	<u>е</u>	е 0	Local x	Backplane	x					_
			0	0	1	x		~	x				
			0	1	0		x	x					_
			0	1	1		х		x				
			1	0	0					х			
			1	0	1						х		
			1	1	0							х	
			1	1	1	Reserved							
12 - 8	BSAB4 - BSAB0						<b>s</b> : These fiv ane) conne		efer to the	e numbe	er of the	data strea	ams
7 - 0 (See Note 1)	BCAB7 - BCAB0		<b>Irce Channel Address Bits</b> : These eight bits refer to the number of the channel that e source (local or backplane) connection.										

#### Table 25 - Backplane Connection Memory Bits

Data Rate	Source Stream	BSAB Bit Usage	BCAB Bit Usage
2 Mb/s	STi0-27	BSAB4-0	BCAB4-0 (32-ch/frame)
4 Mb/s	STi0-18	BSAB4-0	BCAB5-0 (64-ch/frame)
8 Mb/s	STi0-18	BSAB4-0	BCAB6-0 (128-ch/frame)
2-bit subrate	STi0-15	BSAB3-0	BCAB6-0 (128 ch/frame)
4-bit subrate	STi0-15	BSAB3-0	BCAB5 -0 (64 ch/frame)

Table 26 - BSAB and BCAB Bits Usage when Source Streams are from the Local Port

Data Rate	Source Stream	BSAB Bit Usage	BCAB Bit Usage
8 Mb/s	STio0-31	BSAB4-0	BCAB6-0 (128-ch/frame)
16 Mb/s	STio0-15	BSAB3-0	BCAB7-0 (256 ch/frame)

 Table 27 - BSAB and BCAB Bits Usage when Source Streams are from the Backplane Port

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	0	0	0	0	LSRS1	LSRS0

Bit	Name		Description						
15 - 2	Unused	Reserved.	Reserved.						
1 - 0	LSRS1 - LSRS0	Sub-rate Switchin	ng Bits:						
		For the 4-bit wide s	For the 4-bit wide sub-rate switching:						
		LSRS1 - 0	STo Output						
		01	01 Bit 7 - 4 of the 8 bit data						
		00	Bit 3 - 0 of the 8 bit data						
		For 2-bit wide sub-	rate switching:						
		LSRS1 - 0	STo Output						
		11	Bit 7 - 6 of the 8 bit data						
		10	Bit 5 - 4 of the 8 bit data						
		01	Bit 3 - 2 of the 8 bit data						
		00	Bit 1 - 0 of the 8 bit data						

Table 28 - Local Connection Memory High Bits

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
LTM 2	LTM 1	LTM 0	LSAB 4	LSAB 3	LSAB 2	LSAB 1	LSAB 0	LCAB 7	LCAB 6	LCAB 5	LCAB 4	LCAB 3	LCAB 2	LCAB 1	LCAB 0

Bit	Name						Desci	ription							
15 -13	LTM2 - 0	Throughp local ST-B				l Messa	ge Channel	Control	Bits: Th	nese thre	ee bits (	control the			
		<u>LTM2-0</u>				Throu	ughput delay	and Me	ssage N	lode coi	<u>ontrol</u>				
		000	me	mory	is th	e local da	elay from loca ta memory ac out is from the	dress of t	he switch	ned input					
		001 Per-channel constant delay from local interface; the content of the common is the local data memory address of the switched input chan The local ST-BUS output is from the local ST-BUS input.													
		010	010 Per-channel variable delay from backplane interface; the content of the co memory is the backplane data memory address of the switched input char stream. The local ST-BUS output is from the backplane CT-Bus input.												
		011	con	nect	ion m	nemory is	delay from the the backplan he local ST-B	e data me	mory add	dress of t	he switc	hed input			
		100					mode; only the presented to					nection			
		101					ern; the pseuc ut channel.	do randon	n BER tes	st pattern	will be p	presented to			
		110	Pe	r-cha	annel	high-imp	edance. The I	local ST-E	BUS outp	ut is high	-impeda	nce.			
		111 Reserved													
						Inpu	ut Source	Var.	Const.	Msg	BER	Output HiZ			
			LTM2	LTM1	LTM0	Local	Backplane	delay	delay	Mode					
			0	0	0	х		х							
			0	0	1	x			x						
			0	1	0		x	х							
			0	1	1		х		x						
			1	0	0					x					
			1	0	1						х				
			1	1	0			<u> </u>	Deserved			Х			
			1	1	1				Reserved						
	1														

#### Table 29 - Local Connection Memory Low Bits

Bit	Name	Description								
7 - 0 (See Note 1)	LCAB7 - LCAB0	<b>Source Channel Address Bits:</b> These eight bits refer to the number of the channel that is the source (local or backplane) connection.								
Note 1: Only Bits 7-0 will be used for per-channel message mode for the local STo streams.										

#### Table 29 - Local Connection Memory Low Bits (continued)

Data Rate	Source Stream	LSAB Bit Usage	LCAB Bit Usage
8 Mb/s	STio0-31	LSAB4-0	LCAB6-0 (128-ch/frame)
16 Mb/s	STio0-15	LSAB3-0	LCAB7-0 (256 ch/frame)

#### Table 30 - LSAB and LCAB Bits Usage when Source Streams are from the Backplane Port

Data Rate	Source Stream	LSAB Bit Usage	LCAB Bit Usage							
2 Mb/s	STi0-27	LSAB4-0	LCAB4-0 (32-ch/frame)							
4 Mb/s	STi0-18	LSAB4-0	LCAB5-0 (64-ch/frame)							
8 Mb/s	8 Mb/s STi0-18		LCAB6-0 (128-ch/frame)							
2-bit subrate	STi0-15	LSAB3-0	LCAB6-0 (128 ch/frame)							
4-bit subrate STi0-15 LSAB3-0 LCAB5 -0 (64 ch/frame)										
Note: When operating at 4 Mb/s or 8 Mb/s, STo19-27 are driven low.										

Table 31 - LSAB and LCAB Bits Usage when Source Stream are from the Local Port

#### 23.0 DC/AC Electrical Characteristics

#### **Absolute Maximum Ratings\***

	Parameter	Symbol	Min.	Max.	Units
1	Supply Voltage	V <sub>DD</sub>	-0.5	5.0	V
2	BSTio Bias Voltage	V <sub>DD5V</sub>	-0.5	7.0	V
3	Input Voltage	VI	-0.5	V <sub>DD</sub> + 0.5	V
4	Output Voltage	Vo	-0.5	V <sub>DD</sub> + 0.5	V
5	Package power dissipation	PD		2	W
6	Storage temperature	Τ <sub>S</sub>	- 55	+125	°C

\* Exceeding these values may cause permanent damage. Functional operation under these conditions is not implied.

#### $\label{eq:commended_state} \textbf{Recommended Operating Conditions} \ \textbf{-} \ \textbf{Voltages are with respect to ground} \ (\textbf{V}_{SS}) \ \textbf{unless otherwise stated}.$

	Characteristics	Sym.	Min.	Typ.‡	Max.	Units
1	Operating Temperature	T <sub>OP</sub>	-40	25	+85	°C
2	Positive Supply	V <sub>DD</sub>	3.0	3.3	3.6	V
3	BSTio Bias Voltage (3 V PCI Spec)	V <sub>DD5V</sub>	3.0	3.3	3.6	V
3	BSTio Bias Voltage (5 V PCI Spec)	V <sub>DD5V</sub>	4.5	5.0	5.5	V
4	Input Voltage	VI	0		V <sub>DD</sub>	V
5	Input Voltage on 5 V Tolerant Inputs	V <sub>I_5V</sub>	0		V <sub>DD5V</sub>	V

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

#### **DC Electrical Characteristics**<sup>†</sup> - Voltages are with respect to ground (V<sub>ss</sub>) unless otherwise stated.

	Characteristics	Sym.	Min.	Typ.‡	Max.	Units	Test Conditions
1	Supply Current	I <sub>DD</sub>			480	mA	Output unloaded
2	Input High Voltage	V <sub>IH</sub>	0.7V <sub>DD</sub>			V	
3	Input Low Voltage	V <sub>IL</sub>			0.3V <sub>DD</sub>	V	
4	Input Leakage (input pins)	١L			15	μA	$0 < V < V_{DD_{IO}}$
							See Note 1
5	Weak Pullup Current	I <sub>PU</sub>		33	50	μA	Input at 0V
6	Weak Pulldown Current	I <sub>PD</sub>		33	50	μA	Input at V <sub>DD</sub>
7	Input Pin Capacitance	Cl		5	10	pF	
8	Output High Voltage	V <sub>OH</sub>	0.8V <sub>DD</sub>			V	I <sub>OH</sub> = 10 mA
9	Output Low Voltage	V <sub>OL</sub>			0.4	V	I <sub>OL</sub> = 10 mA
10	High Impedance Leakage	I <sub>OZ</sub>			5	μA	0 < V < V <sub>DD_IO</sub>
11	Output Pin Capacitance	CO			15	pF	

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\* Note 1: Maximum leakage on pins (output or I/O pins in high impedance state) is over an applied voltage (Vin).

	Characteristics	Sym.	Level	Units	Conditions
1	CMOS Threshold	V <sub>CT</sub>	0.5V <sub>DD</sub>	V	
2	Rise/Fall Threshold Voltage High	V <sub>HM</sub>	0.7V <sub>DD</sub>	V	
3	Rise/Fall Threshold Voltage Low	$V_{LM}$	0.3V <sub>DD</sub>	V	

#### AC Electrical Characteristics - Timing Parameter Measurement Voltage Levels

#### AC Electrical Characteristics $^{\dagger}$ - Input Frame Pulse and Input Clock Timing

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	FRAME_A_io, FRAME_B_io Input Frame Pulse Width	t <sub>CFPIW</sub>	90	122	180	ns	
2	FRAME_A_io, FRAME_B_io Input Frame Pulse Setup Time	t <sub>CFPIS</sub>	45		90	ns	
3	FRAME_A_io, FRAME_B_io Input Frame Pulse Hold Time	t <sub>CFPIH</sub>	45		90	ns	
4	C8_A_io, C8_B_io Input Clock Period	t <sub>C8MIP</sub>	122-ф		122+ø	ns	
5	C8_A_io, C8_B_io Input Clock High Time	t <sub>C8MIH</sub>	58-ф		64+¢	ns	
6	C8_A_io, C8_B_io Input Clock Low Time	t <sub>C8MIL</sub>	58-ф		64+ <b></b>	ns	
7	Phase Correction	φ	0		10	ns	
8	C8_A_io, C8_B_io Input Rise/Fall Time	t <sub>rC8i</sub> , t <sub>fC8i</sub>	0		5	ns	

† Characteristics are over recommended operating conditions unless otherwise stated.

<sup>‡</sup> Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



Figure 22 - Backplane Frame Pulse Input and Clock Input Timing Diagram

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	Internal Timing Variation	$\Delta^*$	6.50		7.50	ns	
2	Backplane Frame Boundary Offset	<sup>t</sup> FBOS	1.25-∆		<b>2.5+</b> ∆	ns	
3	FRAME_A_io, FRAME_B_io Output Pulse Width	t <sub>CFPOW</sub>	122-∆	122	122+∆	ns	C <sub>L</sub> =30pF
4	Delay from FRAME_A_io, FRAME_B_io output falling edge to C8_a_io,C8_B_io output rising edge	t <sub>CFODF</sub>	61-∆/2		61+∆/2	ns	
5	Delay from C8_A_io,C8_B_io output rising edge to FRAME_A_io,FRAME_B_io output rising edge	t <sub>CFODR</sub>	61-∆/2		61+∆/2	ns	
6	C8_A_io, C8_B_io Output Clock Period	t <sub>C8MP</sub>	122-∆	122	<b>122+</b> ∆	ns	
7	C8_A_io, C8_B_io Output High Time	t <sub>C8MH</sub>	61-∆/2		61+∆/2	ns	C <sub>L</sub> =30pF
8	C8_A_io, C8_B_io Output Low Time	t <sub>C8ML</sub>	61-∆/2		61+∆/2	ns	
9	C8_A_io, C8_B_io Output Rise Time	t <sub>rC80</sub>			13	ns	
10	C8_A_io, C8_B_io Output Fall Time	t <sub>fC80</sub>			14		
11	C32/64o (32.768 MHz) Output Delay Time	t <sub>C32MOD</sub>			Δ	ns	
12	C32/64o (32.768 MHz) Period	t <sub>C32MP</sub>	30.5-∆	30.5	<b>30.5+</b> ∆	ns	C <sub>1</sub> =30pF
13	C32/64o (32.768 MHz) High Time	t <sub>C32MH</sub>	15.25-∆/2		15.25+∆/2	ns	
14	C32/64o (32.768 MHz) Low Time	t <sub>C32ML</sub>	15.25-∆/2		15.25+∆/2	ns	
15	C32/64o (65.536 MHz) Period	t <sub>C64MP</sub>	15.25-∆/2	15.25	15.25+∆/2	ns	
16	C32/64o (65.536 MHz) High Time	t <sub>C64MH</sub>	Δ		0.5 <b>+</b> ∆	ns	
17	C32/64o (65.536 MHz) Low Time	t <sub>C64ML</sub>	Δ		5.5 <b>+</b> ∆	ns	
18	C32/64o Clock Rise Time (32.768 MHz or 65.536 MHz)	t <sub>r320</sub>			5	ns	
19	C32/64o Clock Fall Time (32.768 MHz or 65.536 MHz)	t <sub>f320</sub>			6	ns	

#### AC Electrical Characteristics $^{\dagger}$ - Output Frame Pulse and Output Clock Timing

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\* The AC electrical characteristics are listed as a function of the internal timing variation ( $\Delta$ ) to highlight the value of each parameter independent of the variation. It is important to choose the maximum or minimum  $\Delta$  value correctly for worst case timing calculation. When adding parameters for timing calculation, it is sufficient to include this  $\Delta$  only once in the calculation.





# Figure 23 - Backplane Frame Pulse Output and Clock Output Timing Diagram (in Primary Master Mode and Secondary Master Mode)

#### AC Electrical Characteristics<sup>†</sup> - C20i Master Input Clock Timing

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	C20i Input Clock Period	t <sub>C20MP</sub>	49.995	50	50.005	ns	
2	C20i Input Clock Tolerance		-100		100	ppm	
3	C20i Input Clock High Time	t <sub>C20MH</sub>	20		30	ns	
4	C20i Input Clock Low Time	t <sub>C20ML</sub>	20		30	ns	
5	C20i Input Rise/Fall Time	t <sub>rC20M</sub> , t <sub>fC20M</sub>			10	ns	

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



Figure 24 - Backplane Frame Pulse Input and Clock Input Timing Diagram

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes	
1	CTREF1, CTREF2, LREF0-7 Period	t <sub>R8KP</sub>	121	125	129	μS		
2	CTREF1, CTREF2, LREF0-7 High Time	t <sub>R8kh</sub>	0.09		128.91	μS	8 kHz Mode	
3	CTREF1, CTREF2, LREF0-7 Low Time	t <sub>R8kL</sub>	0.09		128.91	μS	would	
4	CTREF1, CTREF2, LREF0-7 Rise/Fall Time	t <sub>rR8K,</sub> t <sub>fR8K</sub>	0		20	ns		
5	CTREF1, CTREF2, LREF0-7 Period	t <sub>R2MP</sub>	366	488	610	ns		
6	CTREF1, CTREF2, LREF0-7 High Time	t <sub>R2Mh</sub>	90	244	520	ns	2.048 MHz Mode	
7	CTREF1, CTREF2, LREF0-7 Low Time	t <sub>R2ML</sub>	90	244	520	ns	Widde	
8	CTREF1, CTREF2, LREF0-7 Rise/Fall Time	t <sub>rR2M,</sub> t <sub>fR2M</sub>	0		20	ns		
9	CTREF1, CTREF2, LREF0-7 Period	t <sub>R1M5P</sub>	486	648	810	ns		
10	CTREF1, CTREF2, LREF0-7 High Time	t <sub>R1M5h</sub>	90	324	720	ns	1.544 MHz Mode	
11	CTREF1, CTREF2, LREF0-7 Low Time	t <sub>R1M5L</sub>	90	324	720	ns	woue	
12	CTREF1, CTREF2, LREF0-7 Rise/Fall Time	t <sub>rR1M5</sub> , t <sub>fR1M5</sub>	0		20	ns		

#### AC Electrical Characteristics<sup>†</sup> - Reference Input Timing

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



Figure 25 - Reference Input Timing Diagram when the input frequency = 8 kHz



Figure 26 - Reference Input Timing Diagram when the input frequency = 2.048 MHz





	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes	
1	NREFo Output Delay Time	t <sub>ROD</sub>			20	ns		
2	NREFo Clock Period	t <sub>RP</sub>	Same as LR		(DIV1,DIV0) =			
3	NREFo Clock High Time	t <sub>RH</sub>	Same as LR		(0,0)			
4	NREFo Clock Low Time	t <sub>RL</sub>	Same as LR	Same as LREF0-7 Low Time				
5	NREFo Clock Rise/Fall Time	t <sub>rREF</sub> , t <sub>fREF</sub>	0		12 14	ns	in the DOM2 Register	
6	NREFo Clock Period	t <sub>R8KOP</sub>	124.9	125	125.1	μS	(DIV1,DIV0) =	
7	NREFo Clock High Time	t <sub>R8KO2H</sub>	124.4	124.5	124.6	μS	(0,1) or	
8	NREFo Clock Low Time	t <sub>R8KO2L</sub>	<b>488-</b> ∆	488	<b>488+</b> ∆	ns	(DIV1,DIV0) =	
9	NREFo Clock High Time	t <sub>R8KO15H</sub>	124.3	124.4	124.5	μS	(1,0)	
10	NREFo Clock Low Time	t <sub>R8K015L</sub>	648-∆	648	648+∆	ns	in the DOM2 Register	

#### AC Electrical Characteristics<sup>†</sup> - Reference Output Timing

‡ Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



#### Figure 28 - Reference Output Timing Diagram when (DIV1, DIV0) = (0, 0) in DOM2 Register



Figure 29 - Reference Output Timing Diagram when (DIV1, DIV0) = (0, 0) in DOM2 Register



Figure 30 - Reference Input Timing Diagram when (DIV1, DIV0) = (0, 0) in DOM2 Register



Figure 31 - Reference Output Timing Diagram when (DIV1, DIV0) = (1, 0) in DOM2 Register


Figure 32 - Reference Output Timing Diagram when (DIV1, DIV0) = (0, 1) in DOM2 Register

## AC Electrical Characteristics<sup>†</sup> - Local Frame Pulse and Clock Timing, ST\_CKo = 4.096 MHz

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	Local Frame Boundary Offset <sup>1</sup>	t <sub>LFBOS</sub>	Δ		<b>17.5+</b> ∆	ns	
2	ST_FPo0/1 Width	t <sub>FPW4</sub>	<b>244-</b> Δ	244	<b>244+</b> ∆	ns	
3	ST_FPo0/1 Output Delay from Falling edge of ST_FPo0/1 to falling edge of ST_CKo0/1	t <sub>FODF4</sub>	122-∆/2		122+∆/2	ns	C <sub>1</sub> =30pF
4	ST_FPo 0/1Output Delay from Falling edge	t <sub>FODR4</sub>	122-∆/2		122+∆/2	ns	
	of $\overline{ST_CKo0/1}$ to rising edge of $\overline{ST_FPo0/1}$						
5	ST_CKo0/1 Clock Period	t <sub>CP4</sub>	<b>244-</b> ∆	244	<b>244+</b> ∆	ns	
6	ST_CKo0/1 Clock Pulse Width High	t <sub>CH4</sub>	122-∆/2		122+∆/2	ns	
7	ST_CKo0/1 Clock Pulse Width Low	t <sub>CL4</sub>	122-∆/2		122+∆/2	ns	
8	ST_CKo0/1 Clock Rise/Fall Time	t <sub>rC40</sub> , t <sub>C40</sub>			14	ns	

† Characteristics are over recommended operating conditions unless otherwise stated.

<sup>±</sup> Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

Note 1: No jitter presented on input reference clock.



Figure 33 - Local Clock Timing Diagram when ST\_CKo0/1 frequency = 4.096 MHz

		••••=•					
	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	Local Frame Boundary Offset <sup>1</sup>	t <sub>LFBOS</sub>	Δ		17.5 <b>+</b> ∆	ns	
2	ST_FPo0/1 Width	t <sub>FPW8</sub>	<b>122-</b> ∆	122	<b>122+</b> ∆	ns	
3	ST_FPo0/1 Output Delay from Falling edge of ST_FPo0/1 to falling edge of ST_CKo0/1	t <sub>FODF8</sub>	61-∆/2		61+ <u>∆</u> /2	ns	С <sub>L</sub> =30 рF
4	ST_FPo0/1 Output Delay from Falling edge of ST_CKo0-1 to rising edge of ST_FPo0/1	t <sub>FODR8</sub>	61-∆/2		61+ <u>∆</u> /2	ns	
5	ST_CKo0/1 Clock Period	t <sub>CP8</sub>	<b>122-</b> ∆	122	122 <b>+</b> ∆	ns	
6	ST_CKo0/1 Clock Pulse Width High	t <sub>Ch8</sub>	61-∆/2		61+∆/2	ns	
7	ST_CKo0/1 Clock Pulse Width Low	t <sub>CL8</sub>	61-∆/2		61+∆/2	ns	
8	ST_CKo0/1 Clock Rise/Fall Time	t <sub>rC8o</sub> , t <sub>fC8o</sub>			14	ns	

## AC Electrical Characteristics<sup>†</sup> - Local Frame Pulse and Clock Timing, ST\_CK00 = 8.192 MHz

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

Note 1: No jitter presented on input reference clock.



#### Figure 34 - Local Clock Timing Diagram when ST\_CKo0/1 frequency = 8.192 MHz

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	Local Frame Boundary Offset <sup>1</sup>	t <sub>LFBOS</sub>	Δ		17 <b>+</b> ∆	ns	
2	ST_FPo0/1 Width	t <sub>FPw16</sub>	<b>61-</b> Δ	61	61 <b>+</b> ∆	ns	
3	ST_FPo0/1 Output Delay from Falling edge of ST_FPo0/1 to falling edge of ST_CKo0/1	t <sub>FODF16</sub>	30.5-∆/2		30.5+∆/2	ns	
4	ST_FPo Output Delay from Falling edge of	t <sub>FODR16</sub>	30.5-Δ/2		30.5+∆/2	ns	
	ST_CKo0/1 to rising edge of ST_FPo0/1						
5	ST_CKo0/1 Clock Period	t <sub>CP16</sub>	61 <b>-</b> ∆	61	61 <b>+</b> ∆	ns	C <sub>I</sub> =30pF
6	ST_CKo0/1 Clock Pulse Width High	t <sub>Ch16</sub>	30.5-∆/2		30.5+∆/2	ns	- ·
7	ST_CKo0/1 Clock Pulse Width Low	t <sub>CL16</sub>	30.5-∆/2		30.5+∆/2	ns	
8	ST_CKo0/1 Clock Rise/Fall Time	t <sub>rC160</sub> , t <sub>fC160</sub>			14	ns	

## AC Electrical Characteristics<sup>†</sup> - Local Frame Pulse and Clock Timing, ST\_CKo = 16.384 MHz

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

Note 1: No jitter presented on input reference clock.



Figure 35 - Local Clock Timing Diagram when ST\_CKo frequency = 16.384 MHz

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Notes
1	C1M5o Period	t <sub>C1M5oP</sub>	<b>648-</b> ∆	648	648+∆	ns	C <sub>L</sub> =30 pF
2	C1M5o High Time	t <sub>C1M5oH</sub>	324-∆/2	324	324+∆/2	ns	
3	C1M5o Low Time	t <sub>C1M5oL</sub>	324-∆/2	324	324+∆/2	ns	
4	C1M5o Rise Time	t <sub>rC1M50</sub>			10	ns	
5	C1M5o Fall Time	t <sub>fC1M50</sub>			11	ns	

† Characteristics are over recommended operating conditions unless otherwise stated.

<sup>+</sup> Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



Figure 36 - C1M5o Output Clock Timing Diagram

## AC Electrical Characteristics<sup>†</sup>- Output Clock Jitter Generation (Unfiltered)

	Characteristic	Typ.‡	Max.	Units	Notes
1	Jitter at C1M5o (1.544 MHz)	7.4	8.0	ns-pp	Device locks to1.544 MHZ
2	Jitter at ST_CKo0-1 (4.096 MHz)	7.1	8.8	ns-pp	reference input, and no jitter present on the reference
3	Jitter at ST_CKo0-1 (8.192 MHz)	7.0	8.3	ns-pp	present on the relevence
4	Jitter at ST_CKo0-1 (16.384 MHz)	7.6	9.9	ns-pp	

† Characteristics are over recommended operating conditions unless otherwise stated.

<sup>‡</sup> Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

	Characteristic	Sym.	Min.‡	Typ.‡	Max.	Units	Test Conditions
1	STio0-31 Input Data Sample Point	t <sub>SAMP8</sub>	91.5	91.5	91.5	ns	
2	STio0-31 Input Setup Time	t <sub>CIS8</sub>	5+∆			ns	
3	STio0-31 Input Hold Time	t <sub>CIH8</sub>	<b>5+</b> ∆			ns	
4	STio0-31 Output Delay Active to Active	t <sub>DOD8</sub>	1-∆		5+∆	ns	C <sub>L</sub> = 30 pF, Note 1
5	Per Channel boundary HiZ	t <sub>DOZ8</sub> t <sub>ZDO8</sub>		10 10		ns ns	R <sub>L</sub> =1K, C <sub>L</sub> =30 pF, Note 2

#### AC Electrical Characteristics<sup>†</sup> - Backplane Serial Streams with Data Rate of 8 Mb/s

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\* Note 1: To meet the H.110 output timing requirement, the output delay time can be reduced further by programming the backplane output advancement registers (BOA0 - 3).

\* Note 2: High Impedance is measured by pulling to the appropriate rail with R<sub>L</sub>, with timing corrected to cancel the time taken to discharge C<sub>L</sub>.



Figure 37 - Backplane Serial Stream Timing when the Data Rate is 8 Mb/s

	Characteristic	Sym.	Min.	Тур.‡	Max.	Units	Test Conditions
1	STio0-15 Input Data Sample Point	t <sub>SAMP16</sub>	46	46	46	ns	
2	STio0-15 Input Setup Time	t <sub>CIS16</sub>	<b>5+</b> ∆			ns	
3	STio0-15 Input Hold Time	t <sub>CIH16</sub>	<b>5+</b> ∆			ns	
4	STio0-15 Output Delay	t <sub>DOD16</sub>	<b>1-</b> ∆		<b>5+</b> ∆	ns	C <sub>L</sub> = 30 pF
	Active to Active						

AC Electrical Characteristics<sup>†</sup> - Backplane Serial Streams with Data Rate of 16 Mb/s

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.



Figure 38 - Backplane Serial Stream Timing when the Data Rate is 16 Mb/s

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Test Conditions
1	STo Delay - Active to Active						
	@2.048 Mb/s	t <sub>SOD2</sub>	-10- $\Delta^{*}$		1.5- $\Delta^*$	ns	C <sub>L</sub> = 30 pF
	@4.096 Mb/s	t <sub>SOD4</sub>	-10- $\Delta^{*}$		1.5- $\Delta^*$	ns	C <sub>L</sub> = 30 pF
	@8.192 Mb/s	t <sub>SOD8</sub>	-10- $\Delta^{*}$		1.5-∆ <sup>*</sup>	ns	C <sub>L</sub> = 30 pF

AC Electrical Characteristics<sup>†</sup> - Local Serial Stream Output Timing

† Characteristics are over recommended operating conditions unless otherwise stated.

<sup>‡</sup> Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\* See Section 7.0, "Local Output Timing Considerations" on page 21



Figure 39 - Local Serial Stream Output Timing

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Test Conditions
1	STi Input Data Sample Point						
	@2.048 Mb/s	t <sub>SAMP2L</sub>	366	366	366	ns	
	@4.096 Mb/s	t <sub>SAMP4L</sub>	183	183	183	ns	
	@8.192 Mb/s	t <sub>SAMP8L</sub>	91.5	91.5	91.5	ns	
2	STi Setup Time						
	@2.048 Mb/s	t <sub>SIS2</sub>	5+∆			ns	
	@4.096 Mb/s	t <sub>SIS4</sub>	5+ $\Delta$			ns	
	@8.192 Mb/s	t <sub>SIS8</sub>	5+∆			ns	
3	STi Hold Time						
	@2.048 Mb/s	t <sub>SIH2</sub>	5+∆			ns	
	@4.096 Mb/s	t <sub>SIH4</sub>	5+∆			ns	
	@8.192 Mb/s	t <sub>SHI8</sub>	<b>5+</b> ∆			ns	



Figure 40 - Local Serial Stream Input Timing

	Characteristic	Sym.	Min.	Typ.‡	Max.	Units	Test Conditions
1	STo/STio Delay - Active to High-Z - High-Z to Active 2.048 Mb/s (local) 4.096 Mb/s (local) 8.192 Mb/s (local) 8.192 Mb/s (backplane) 16.384 Mb/s (backplane)	t <sub>DZ,</sub> t <sub>ZD</sub>	-12-Δ -12-Δ -12-Δ -1-Δ -1-Δ		3.5-Δ 3.5-Δ 3.5-Δ 7+Δ 7+Δ	ns ns ns ns ns	R <sub>L</sub> =1K, C <sub>L</sub> =30pF, See Note 1.
2	Output Driver Enable (ODE) Delay - High-Z to Active 2.048 Mb/s (local) 4.096 Mb/s (local) 8.192 Mb/s (local) 8.192 Mb/s (backplane) 16.384 Mb/s (backplane)	t <sub>ZD_ODE</sub>			37 37 37 20 20	ns ns ns ns ns	
2	Output Driver Disable (ODE) Delay - Active to High-Z 2.048 Mb/s (local) 4.096 Mb/s (local) 8.192 Mb/s (local) 8.192 Mb/s (backplane) 16.384 Mb/s (backplane)	t <sub>DZ_ODE</sub>			20 20 20 20 20	ns ns ns ns ns	

## AC Electrical Characteristics $^{\dagger}$ - Local and Backplane Tristate Timing

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\* Note 1: High Impedance is measured by pulling to the appropriate rail with R<sub>L</sub>, with timing corrected to cancel the time taken to discharge C<sub>L</sub>.



Figure 41 - Local Serial Output and External Control



Figure 42 - Backplane Serial Output and External Control



Figure 43 - Output Driver Enable (ODE)

	Characteristics	Sym.	Min.	Typ.‡	Max.	Units	Test Conditions <sup>1</sup>
1	CS setup from DS falling	t <sub>CSS</sub>	0			ns	
2	R/W setup from DS falling	t <sub>RWS</sub>	15			ns	
3	Address setup from DS falling	t <sub>ADS</sub>	5			ns	
4	CS hold after DS rising	t <sub>CSH</sub>	0			ns	
5	R/W hold after DS rising	t <sub>RWH</sub>	0			ns	
6	Address hold after DS rising	t <sub>ADH</sub>	5			ns	
7	Data setup from DTA Low on Read	t <sub>DDR</sub>	20			ns	C <sub>L</sub> =30pF
8	Data hold on read	t <sub>DHR</sub>			20	ns	C <sub>L</sub> =30pF, R <sub>L</sub> =1K See Note 2
9	Valid Write Data Setup	t <sub>WDS</sub>			20	ns	
10	Data hold on write	t <sub>DHW</sub>	8			ns	
11	Acknowledgment Delay: Reading/Writing Registers Reading/Writing Memory	t <sub>AKD</sub>		97/82 158/114	110/95 171/127	ns ns	C <sub>L</sub> =30pF C <sub>L</sub> =30pF
12	Acknowledgment Hold Time	t <sub>AKH</sub>			30	ns	C <sub>L</sub> =30pF, R <sub>L</sub> =1K, See Note 2

#### AC Electrical Characteristics<sup>†</sup> - Motorola Non-Multiplexed Bus Mode

† Characteristics are over recommended operating conditions unless otherwise stated.

‡ Typical figures are at 25°C and are for design aid only: not guaranteed and not subject to production testing.

\*Note 1: A delay of 100 microseconds must be applied before the first microprocessor access is performed after the RESET pin is set high.

\*Note 2: High Impedance is measured by pulling to the appropriate rail with R<sub>L</sub>, with timing corrected to cancel the time taken to discharge C<sub>L</sub>.





	Characteristic	Sym.	Min.	Тур.	Max.	Units	Notes
1	TCK Clock Period	t <sub>TCKP</sub>	200			ns	
2	TCK Clock Pulse Width High	t <sub>TCKH</sub>	80			ns	
3	TCK Clock Pulse Width Low	t <sub>TCKL</sub>	80			ns	
4	TMS Set-up Time	t <sub>TMSS</sub>	10			ns	
5	TMS Hold Time	t <sub>TMSH</sub>	10			ns	
6	TDi Input Set-up Time	t <sub>TDIS</sub>	20			ns	
7	TDi Input Hold Time	t <sub>TDIH</sub>	20			ns	
8	TDo Output Delay	t <sub>TDOD</sub>			30	ns	C <sub>L</sub> =30pF
9	TRST pulse width	t <sub>TRSTW</sub>	20			ns	C <sub>L</sub> =30pF
10	Reset pulse width	t <sub>RSTW</sub>	400			ns	C <sub>L</sub> =30pF

## AC Electrical Characteristics<sup>†</sup> - JTAG Test Port and Reset Pin Timing

† Characteristics are over recommended operating conditions unless otherwise stated.



## Figure 45 - JTAG Test Port Timing Diagram



#### Figure 46 - Reset Pin Timing Diagram

## 24.0 Trademarks

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